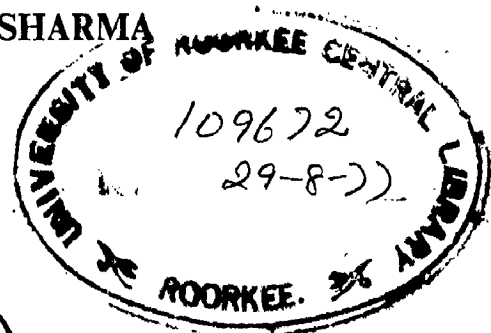


DESIGN OF NUCLEAR REACTOR CONTROL SYSTEM WITH TIGHT FEEDBACK

A DISSERTATION
submitted in partial fulfilment
of the requirements for the award of the Degree
of
MASTER OF ENGINEERING
in
ELECTRICAL ENGINEERING
(System Engineering & Operations Research)

By

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ROORKEE, U.P.
November, 1976

C E R T I F I C A T E

Certified that the dissertation entitled
" DESIGN OF NUCLEAR REACTOR CONTROL SYSTEM WITH TIGHT
FEEDBACK" which is being submitted by Sri Suresh Kumar
Sharma in partial fulfillment for the award of Degree
of Master of Engineering in " SYSTEM ENGINEERING AND
OPERATIONS RESEARCH" of the University of Roorkee,
Roorkee is a record of student's own work carried out
by him under my supervision and guidance. The matter
embodied in this dissertation has not been submitted
for the award of any other degree or diploma.

This is further to certify that he has worked
for a period of *Eight* months from *March* to *Nov.*, 1976
for preparing dissertation for Master of Engineering
Degree at this University.

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SYNOPSIS

The present work covers the design of Nuclear Reactor Control System with tight feedback. The design uses the modified form of the Sequential Return Difference (SRD) method. The SRD method is modified to reduce computational work and, hence, make the algorithm faster.

Sequential return difference method gives a satisfactory closed-loop linear system via a sequence of single loop designs. The stability of the resulting system may be checked by classical techniques such as Routh Hurwitz Criterion, Nyquist Criterion or Root locii. Here, as the number of equations to be checked for stability is very large, Routh Hurwitz criterion is utilized. In the 1st Chapter return difference and return ratio matrices are discussed, and how these are useful in the synthesis is given as a background for the SRD method. Second Chapter describes the SRD method. Modified transfer function matrix achieved from the SRD method is checked for step input. All the output are given in the end.

The SRD method is found to give good controller designs using only single-input single-output classical (well established) design techniques in the frequency domain. This powerful method can be fruitfully used to design complex systems if inter-active computer graphic terminals are available. In this work, however, only IBM 1620 with punched output facility was available and

inordinately long CPU time (this computer is very slow-
50 times slower than IBM 360) was involved.

Computer programs in FORTRAN II were developed
for

- (a) Sequential Return difference algorithm
- (b) Routh Hurwitz criterion (**SUBROUTINE**)
- (c) Ensuring tight feedback
- (d) Nyquist criterion
- (e) Step response with subroutines of Polynomial
multiplication and polynomial summation
- (f) Determining the Laplace inverse and generation
of time response curves for three output.

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CHAPTER-I

INTRODUCTION

Design of Control Systems has seen rapid progress in recent years. With modern developments in design techniques, complexities in the controller structure are also increasing at the same rate. For this reason, designers now have started to look for design of multivariable systems using the old classical techniques. It has also been brought up to the practical stage by few designers. The present work is based upon the views of such control Engineers. The work of a control engineer is not to just formulate and analyse the control problem but ultimately he has to be involved in the design of total control systems which is a complex subject in itself.

A design problem has never a unique solution. A given set of specifications may be satisfied by a number of transfer functions. More-over leaving technical specifications, the requirements of weight, size and cost may introduce extra constraints. Here the considerations for weight, size and cost has not been taken into account. The problem has been tackled only from stability and performance point of view.

1.1. NEED OF THE CONTROL SYSTEM:

In a feedback control system, the requirement of a controller can in no case be avoided. Whenever the error signal between desired and actual output is received,

one should be able to control the system to make the error to converge to zero eventually. In the case of a Nuclear Reactor on sudden removal of power, fluctuations will occur at the outputs such as voltage, frequency and power. These fluctuations have to be damped out in a reasonable time so that relevant safety regulations are adhered to. Therefore controlling the output of a reactor under required specifications becomes a necessity.

1.2. TOWARDS THE DESIGN OF CONTROL SYSTEM :

The last few decades has seen an explosion of knowledge in the field of control systems. Two distinct classifications may be made viz classical control methods and modern control methods. The classical methods, which allow more insight in the dynamics of the system in the form of root locus. Bode plots, Nyquist and Polar plots etc., however, suffer from the fact that they are basically suited to single-input single-output systems only, The modern control theory, on the other hand, gives powerful tools for the design of controllers for multi-variable systems. Barring the new techniques of pole placement etc. modern control methods eg optimal control etc. suffer from the basic drawback of requirement of on line digital controller for control purpose plus the added disadvantage of trial-and-error schemes involved in selecting the weighting matrices.

With the advent of high speed digital computers, methods came up which utilize the classical well-known (well understood) techniques eg. Nyquist Bode etc. for the design of multivariable systems. The work of H.H. Rosenbrock and D.Q. Mayne falls under this class of classico-modern approach.

1.3. DESIGN CRITERIONS:

Besides providing a useful theoretical basis, ROSENBROCK described a useful technique based on the **inverse** Nyquist array, for designing linear multivariable systems, and a similar technique using the Nyquist array and classical methods are extended by MacFarlane. In this approach extensions are done by Dr. D.Q. Mayne. Here a series compensator is chosen to transform the plant transfer function $G_P(S)$, which is the $m \times m$ matrix transfer function of the system being controlled to $G(S) = G_P(S) G_C(S)$ where $G(S)$ is diagonally dominant. Then m single loop control problems are considered, choosing $K_i(S)$, $i = 1-m$.

The design criterions are assumed to be :

- i) Performance
- ii) Stability
- iii) Security or integrity, the maintenance of stability in face of component failure
- iv) Low interaction.

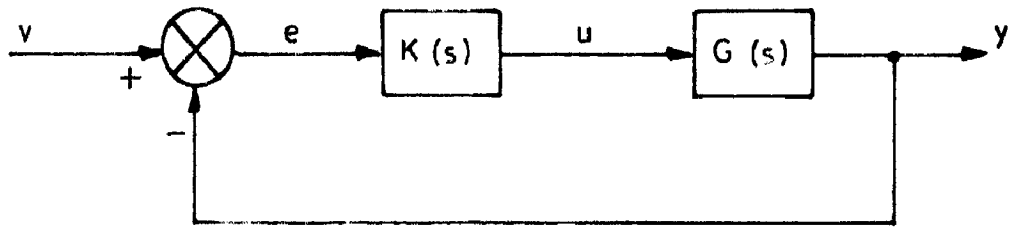


FIG.1.1 MULTIVARIABLE CONTROL SYSTEM.

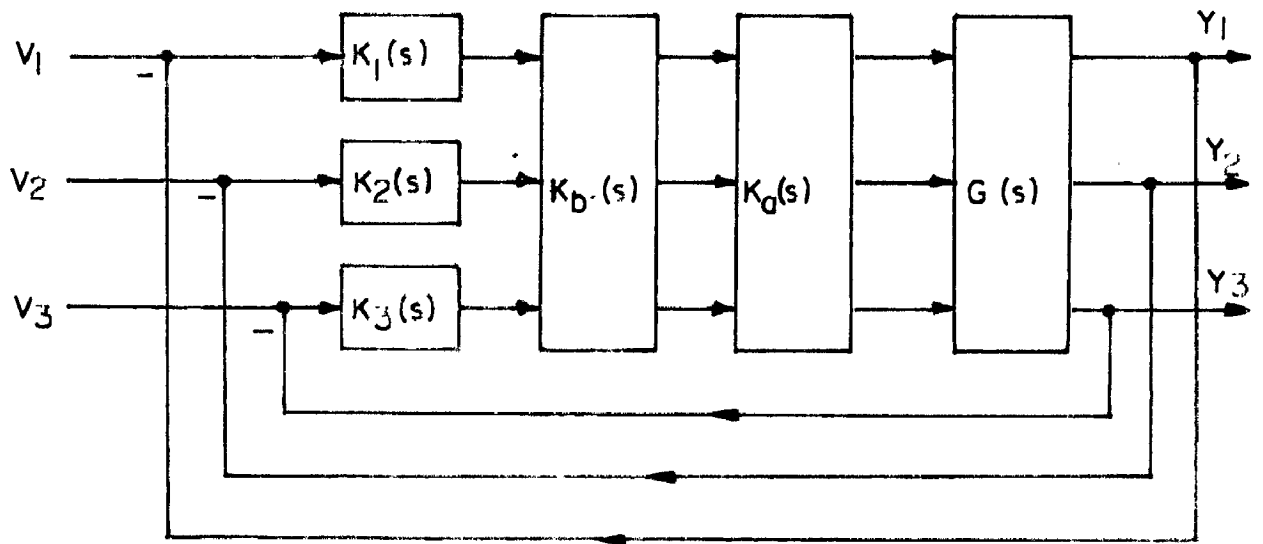


FIG.1.2 STRUCTURE OF THE CONTROLLER.

Stability and performance are basic criteria. Security is achieved in practice by a variety of means e.g. switching to alternative controller. If any component fails, stability may have to be achieved at the expense of performance. Hence a design method which ensures stability in the event of the failure of any specified combinations of N components, should also be flexible enough to include the case $N = 0$. Last is interaction which at low frequencies is automatically reduced in high performance systems, and may not be important at high frequencies, so that, like security, it may not be an important factor in some designs. Thus in the present case the criterias for stability and performance have only been taken into consideration.

1.4. STRUCTURE OF THE CONTROLLER:

The plant has the $n \times n$ transfer function matrix $G(S)$, and the controller is represented by an $n \times n$ matrix $K(S)$. The object is to find a suitable matrix $K(S)$ which will ensure that the closed loop system meets certain performance specifications.

It will be assumed that the elements of $G(S)$ and $K(S)$ are rational polynomial functions of S , and that neither $G(S)$ nor $K(S)$ is identically zero. It will also be assumed that all the zeros of $K(S)$ are in the open left half plane, because the right half plane zeros in $K(S)G(S)$ give rise to control difficulties, so that

there will be no incentive to introduce them in $K(S)$. Finally it is assumed that the plant from which $G(S)$ arises is asymptotically stable before control is applied, and that $K(S)$ has all its poles in the open left half plane.

Since the objective is to design a suitable controller $K(S)$, it is desirable to know what structure is adequate to describe a general $K(S)$. Any such $K(S)$ can be written as a product.

$$K(S) = K_a K_b(S) K_c(S)$$

where three matrices K_a , $K_b(S)$ and $K_c(S)$ has the following properties.

The matrix K_a is a permutation matrix. It therefore represents a preliminary renumbering of the inputs to $G(S)$, which usually will be done so that the new input i affects chiefly the output i .

The matrix $K_b(S)$ has determinant $K_b(S) = 1$ and represents a sequence of elementary column operations. Each such operation consists of adding, to column j of $Q(S)$ operated on, a multiple of $d_{ij}(S)$ by column i . Here $d_{ij}(S)$ is a rational polynomial having all its zeros in left half plane. And where $Q(S) = G(S) K(S)$, $K_b(S)$ is used only to make the plant transfer function diagonally dominant or it accomplishes a modification of the interaction in the plant.

The matrix $K_c(S)$ is diagonal and its nonzero entries have all their poles and zeros in the open left

half plane. $K_c(S)$ may be written as

$$K_c(S) = \text{diag} (K_1 (S))$$

$K_c(S)$ represents m independent controllers. The m loops which contain the $K_i(S)$ will be called the m principal loops. The importance of the decomposition of $K(S)$ into K_a , K_b and K_c is that the successive application of K_a , K_b and K_c is sufficient to generate the most general K satisfying the conditions on K . The structure is given in fig.1.2.

In fig.1.2 K_1 , K_2 and K_3 are 3 independent controllers. $K_b(S)$ is to make $G(S)$ diagonally dominant and K_a is for renumbering the inputs.

1.5. SRD AND INA METHODS: A COMPARISON :

Inverse Nyquist Array method for the design of linear multivariable systems was given by H.H. ROSENBROCK in his pioneering paper (2). This method was perhaps the first to design the multivariable systems with classical techniques having very much satisfactory results. In this method, it is necessary to make the plant transfer function diagonally dominant. Sequential return difference method given by Dr. D.J. Mayne has some modifications over the work of ROSENBROCK and has generated a new algorithm for the design of multivariable systems. SRD method has its supremacy in the sense that diagonally dominance is not necessarily needed. Thus for employing the SRD method

one is not supposed to choose $K_p(S)$ shown in fig.2.1. K_a is dependent of number of loops closed. Thus SRD method needs only the proper selection of $K_i(S)$.

The distinguishing feature of SRD method is that it calculates at the i th iteration, the exact modified transfer function of the system with the previous $i-1$ loops closed. If stability and disturbance attention are the sole design requirements, the multivariable controller can be synthesised by this method with good results. In SRD method compensator may have to be designed, if any diagonal element of the modified transfer function matrix is nonminimal phase i.e. it has zeros in the right half plane. Compensator is to cancel the zeros of right half plane, and the determinant of compensating matrix must be unity.

CHAPTER - II

RETURN DIFFERENCE AND RETURN RATIO MATRICES AND THEIR UTILITY IN THE DESIGN OF CONTROL SYSTEMS

The concepts of return difference and return ratio are shown to play a fundamental role in the analysis and synthesis of multivariable feedback control systems. Matrix transfer functions are regarded as operators over the field of rational functions in the complex variables. The eigen values of such operators are identified as characteristic transfer functions. The corresponding characteristic frequency responses provide a simple and natural link between classical single loop design techniques and multivariable system feedback theory. These concepts serve as a unifying thread in a coherent and systematic discussion of multivariable feedback system design techniques. Moreover these concepts have made the analysis and synthesis very simple.

2.1. RETURN DIFFERENCE MATRIX

From fig. 2.1:

$r(S)$ = $m \times 1$ matrix of reference input transforms

$e(S)$ = $m \times 1$ matrix of error transforms

$y(S)$ = $m \times 1$ matrix of plant output transforms

$u(S)$ = $r \times 1$ matrix of plant input transforms

$K(S)$ = $r \times m$ matrix of controller transfer functions

$G(S)$ = $m \times r$ matrix of plant transfer functions

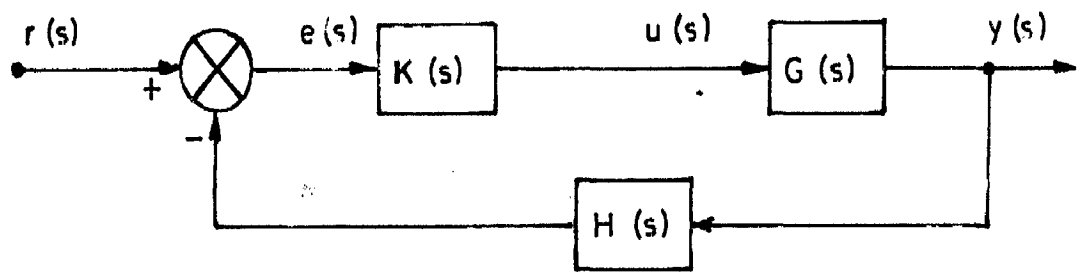


FIG. 2.1

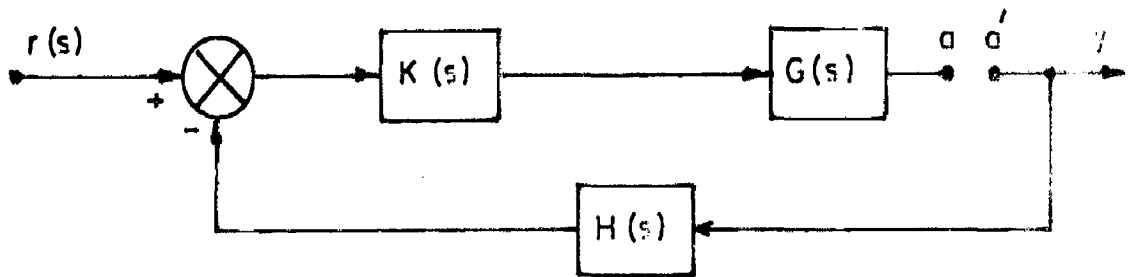


FIG. 2.2

$K(S)$, $G(S)$ and $H(S)$ are matrices over the field of rational functions in the complex variable S . The closed loop system transfer function matrix is given by

$$R(S) = (I_m + G(S) K(S) H(S))^{-1} G(S) K(S) \dots (2.1)$$

Now suppose all the feedback loops are broken as shown in fig.2.2. and a signal transform vector $a(S)$ is injected at point a . The transform of the signal returned at a' is then

$$- G(S) K(S) H(S) a(S)$$

and the difference between injected and returned signals is thus

$$\{I_m + G(S) K(S) H(S)\} a(S) = T(S) a(S) \dots (2.2)$$

where $T(S) = \{I_m + G(S) K(S) H(S)\} \dots (2.3)$

$T(S)$ is defined as the system return difference matrix.

2.2. RETURN RATIO MATRIX :

The matrix

$$F(S) = G(S) K(S) H(S) \dots (2.4)$$

is defined as the system return ratio matrix, so that we have

$$T(S) = I_m + F(S) \dots (2.5)$$

Here $T(S)$ and $F(S)$ both are natural generalisations of the equivalent scalar concept introduced by Bode. Using arguments given by ROSENBROCK, it can be shown that

Let

$$T(S) = \frac{\text{Closed-loop characteristic Polynomial}}{\text{Open-loop characteristic Polynomial}} \dots(2.6)$$

The proof of (2.6) is given in the appendix of (4). This is the fundamental equation relating open and closed loop behaviour in multiple loop control systems.

2.3. STABILITY IN TERMS OF RETURN DIFFERENCE MATRIX FOR MULTIPLE LOOP SYSTEMS:

Assume that the system is open loop stable. The open loop characteristic polynomial will then have no zeros in the closed right half complex plane. Thus it follows from equation (2.6) that the closed loop characteristic polynomial will not vanish in the closed right half complex plane if, and only if, $\det T(S)$ does not vanish in the closed right half complex plane. Thus it is sufficient to check only $T(S)$ instead of closed loop characteristic polynomial.

2.3.1. Nyquist Type of Criterion:

Let D be a contour in the complex plane consisting of the imaginary axis from $-ja$ to $+ja$ and a semicircle centred at the origin of radius a in the right half plane. Further, let a be large enough to ensure that every zero and pole of $\det GK(S)$ and $\det R(S)$ which is in the open right half plane lies within D .

Suppose D maps into a closed curve T in the complex under the mapping $\det T(S)$.

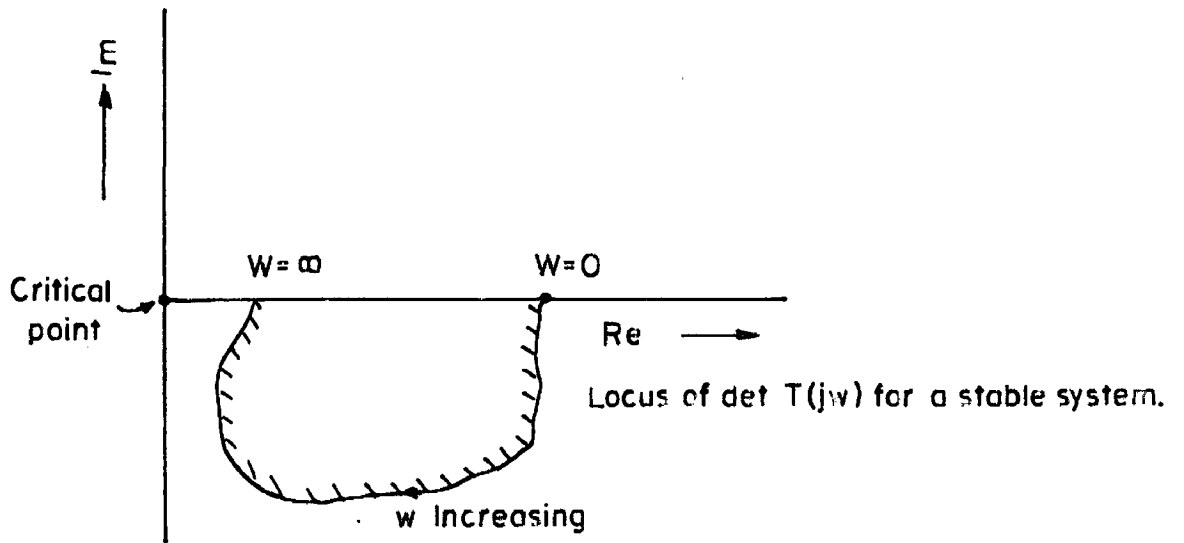


FIG.2.3 SIMPLE MULTIVARIABLE NYQUIST CRITERION

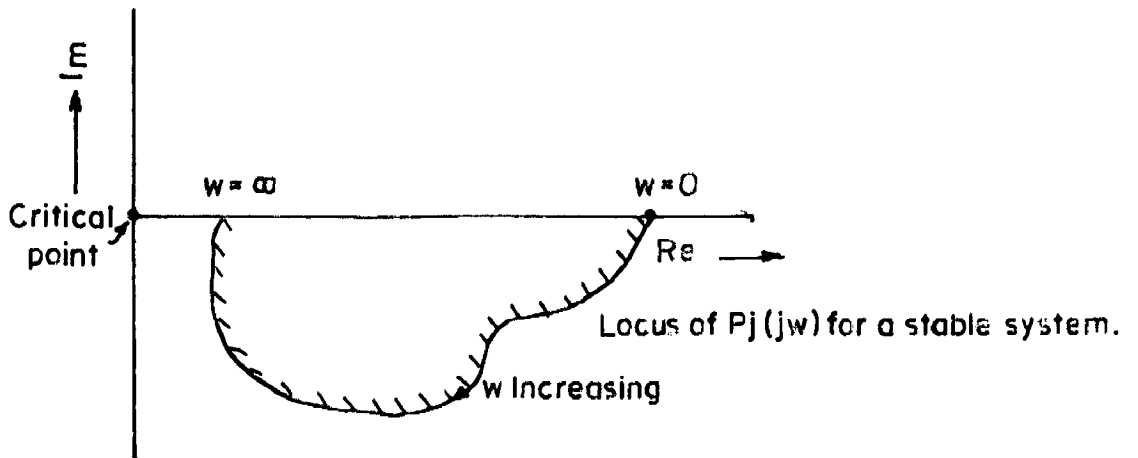


FIG.2.4 EXTENDED NYQUIST CRITERION FOR CHARACTERISTIC FREQUENCY RESPONSE.

Thus the system is closed loop stable if no point within D maps on the origin of the complex plane under the mapping $\det T(S)$.

Thus the system is closed loop stable if T does not enclose the origin of the complex plane. If $\det F(S) \rightarrow 1$ as $S \rightarrow \infty$, then, taking D as arbitrarily large, we can conveniently refer to T as the locus $\det F(j\omega)$. This gives the multiple-loop Nyquist type of criterion for stability, shown in fig.2.3.

Let the eigen values of $T(S)$ be $P_j(S)$. $j=1 \dots m$ we then have that

$$\det T(S) = \prod_{j=1}^m P_j(S) \quad \dots \quad (2.7)$$

therefore, $\det T(S)$ will not vanish for any S enclosed by D if none of $P_j(S)$; $j = 1 \dots m$ vanish for any S enclosed by D . Let D maps into D_j in the complex plane under $P_j(S)$; $j = 1 \dots m$, then for a stable system T will not enclose the origin of the complex plane for $j=1 \dots m$. Thus the system will be stable with all loops if none of enclose the origin of the complex plane for $j=1 \dots m$.

Fundamental stability property of complex-plane loci of the return difference matrix eigen values can be stated as : The system is closed loop stable if all the eigen value loci $P_j(j\omega)$ for $j=1 \dots m$ satisfy the Nyquist criterion as illustrated in fig.2.4.

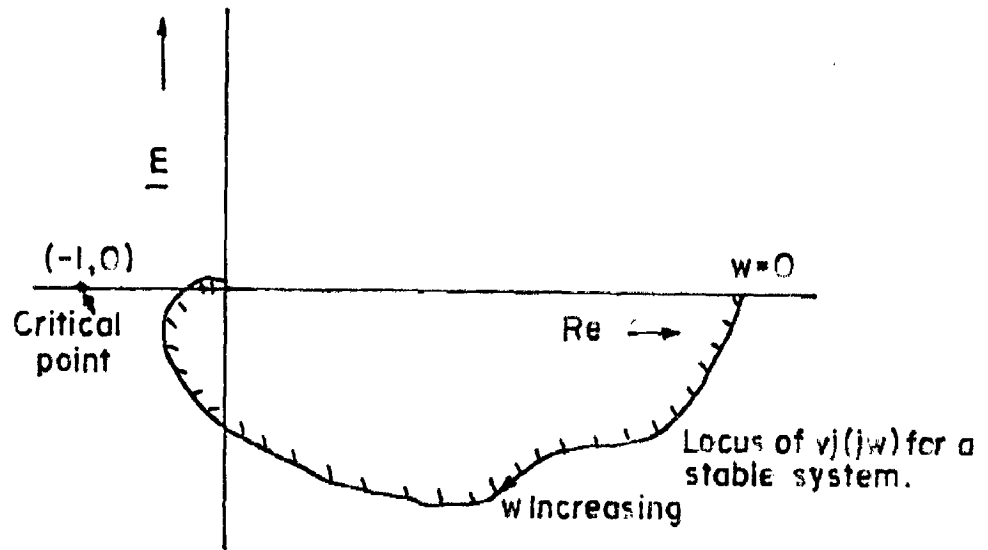


FIG.2.5 CHANGE IN CRITICAL POINT.

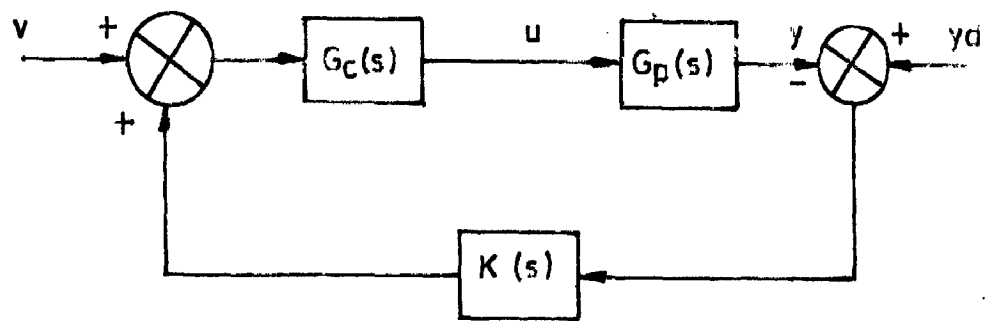


FIG.3.1 MULTIVARIABLE CONTROL SYSTEM.

This criterion can equally well be stated in terms of the return ratio matrix. Since

$$T(S) = I_m + F(S)$$

The eigen values of $T(S)$ and $F(S)$ are simply related via the eigen value shift theorem. This shows that if

$V_j(S) : j = 1 \dots m$ are the eigen values of $F(S)$, then

$$P_j = 1 + V_j(S) \quad j = 1 \dots m \quad (2.8)$$

In terms of the return ratio matrix, therefore, we simply get a unit shift in the location of the critical point. The system is closed loop stable if all the eigen value loci $V_j(j\omega)$ for $j=1, \dots, m$ satisfy the Nyquist criterion as in fig.2.5.

The consideration of the behaviour of $\det T(S)$ along the imaginary axis in the complex plane shows that the old criterion for scalar systems is simply extended to the multiple loop case.

2.3.2. Routh Hurwitz Criterion :

In the case of return difference and return ratio matrix, the application of Routh Hurwitz Criterion is very simple. Here if scalar return difference is known before hands, the stability may be checked by the direct application of the criterion for each scalar return difference separately. On the other hands if scalar return ratio is known then the closed loop stability may be decided on the basis of

$$t_j(S) = 1 + f_j(S) \quad \dots (2.9)$$

Where t_j = Scalar return difference

f_j = Scalar return ratio

This is useful, when it is difficult to use the Nyquist type of criterion because of computation difficulties.

CHAPTER - III

SEQUENTIAL RETURN DIFFERENCE METHOD FOR THE DESIGN OF LINEAR MULTIVARIABLE SYSTEMS:

The classical frequency methods for designing single loop control systems have proved to be so useful that it is surprising that so little effort has been devoted to extending these techniques to multivariable systems. This may be due to the development of modern control theory, which though originally motivated by open-loop trajectory optimization problems, yielded useful and elegant results for linear multivariable control and filtering problems. The resultant controllers are complex, however, requiring a dynamic filter or observer of almost the same order of complexity as the plant or process being controlled. In order to reduce the complexity of controller ROSENBROCK(2) redried attention to the problem of extending classical procedures to multivariable problems. ROSENBROCK described a useful technique using the Inverse Nyquist Array, for designing multivariable systems. These methods are extended by MACFARLANE (4). Based on the work of (4) a method named as sequential return difference is given by Mayne (6).

Sequential return difference method gives a satisfactory closed- loop linear system via a sequence of single loop designs, in which classical techniques such as Nyquist diagram, root loci, Routh's Criterion etc. are employed.

Stability and performance are the basic criteria for the design. Security is achieved in practice by a variety of means e.g. switching to alternative controller. In the case of component failure, security may have to be achieved at the expense of performance. Interaction at low frequencies is automatically reduced in high performance systems, and may not be important at high frequencies so that, like security it may not be an important factor in some designs. Hence in the sequel, a basic design algorithm, consisting of a sequence of single loop designs, for achieving good performance and stability is described. In this case diagonal dominance is not necessarily required, though it can be employed if it is desired so that increased flexibility in the choice of the compensating matrix $G_c(S)$ is available. Diagonal dominance (2) however automatically provides security against arbitrary, output transducer failure (6) and also limits interaction.

3.1. ASSUMPTIONS AND PROOFS OF RESULTS TO BE USED IN GENERATING THE SRD ALGORITHM :

The system considered is shown in the fig.3.1
The process to be controlled has $m \times m$ transfer function $G_p(S)$. $G_c(S)$ is a $m \times m$ compensator matrix and :

$$G(S) = G_p(S) G_c(S) \dots \quad (3.1)$$

$K(S)$ is a $m \times m$ diagonal matrix. Assumptions made are :

(A1) Neither $G(S)$ nor $K(S)$ are identically zero.

The matrix return difference $T(S)$ is defined to be

$$T(S) = I_n + G(S) K(S) \quad \dots (3.2)$$

The system has a state space representation :

$$\dot{x}(t) = AX(t) + Be(t) \quad \dots (3.3)$$

$$y(t) = CX(t) \quad \dots (3.4)$$

$$e(t) = y_d(t) - y(t) \quad \dots (3.5)$$

$$\text{i.e. } \dot{X}(t) = (A - BC) X(t) + B Y_d(t) \quad \dots (3.6)$$

The expressions

$$W_o(S) = SI - A \quad \dots (3.7)$$

$$W_c(S) = SI - A + BC \quad \dots (3.8)$$

are respectively the open loop and closed loop characteristic polynomials.

$$\text{Let } \bar{G}(S) = G(S) K(S) \quad \dots (3.9)$$

the loop transfer function, and let $R(S)$ denote the closed loop transfer function relating $y(s)$ to $Y_d(S)$. Clearly

$$\bar{G}(S) = C (SI - A)^{-1} B \quad \dots (3.10)$$

since

$$T(S) \bar{G}(S) = \bar{G}(S) T(S) \quad \dots (3.11)$$

We have

$$R(S) = T^{-1}(S) \bar{G}(S) = \bar{G}(S) T^{-1}(S) \quad \dots (3.12)$$

$W_o(S)$ and $W_c(S)$ are related by the following well known result (4).

$$T(S) = W_c(S) / W_o(S) \quad \dots (3.13)$$

Now we make the further assumption.

(A2) The, process, with transfer function $G_p(S)$, is open loop asymptotically stable. Note that the assumption of asymptotic stability is made for simplicity in presentation.

(A3) It is also assumed that $G_o(S)$ and $K(S)$ have poles and zeros in the left half plane only and $|G_o(S)| = 1$ then $W_o(S)$ has zeros in the open left half plane only.

From equation (3.13) it follows that closed loop system is asymptotically stable if and only if the locus T_m of $T(j\omega)$ does not encircle or pass through the origin. And for reducing interaction, high gain loops are required so as

$$T^{-1}(j\omega) \rightarrow 1 \quad \text{for } \omega \rightarrow \infty \quad \dots(3.14)$$

For convenience some extra terms appropriate to the condition when the first j loops are closed and the remaining open, i.e. $K_j(S) = 0$.

$j = j+1 \dots m$, are defined

$$K_i(S) = \text{diag} (k_1(S), \dots, k_j(S), 0, 0, \dots) \quad (3.15)$$

$$T_i(S) = I_m + G(S) K_i(S) \quad (3.16)$$

$$G_i(S) = T^{-1}(S) G(S) \quad (3.17)$$

for $i = 0 \dots m$ clearly $T_0(S) = I_m$, $G_0(S) = G(S)$

$G_i(S)$ is the transfer function relating y to v in fig.3.1. when loop $j+1 \dots m$ are open, the scalar return difference

$$t_i(S) = 1 + k_j(S) g_{ij}(S) \quad (3.18)$$

where $g_{ij}(S)$ is the ij th element of $G^k(S)$.

Let $g_{i\cdot}, g_{\cdot i}$ denote respectively, the i th row and i th column of G . Let S_0 denote the open loop state space representation (A, B, C) of $G_P(S) G_C(S) K(S)$ given in equation (3.3) and (3.4). Let S_m denote the closed loop state space representation $(A-BC, B, C)$ given in equations (3.3), (3.4) and (3.5). For $i=1\dots m$, let S_i denote the state space representation $(A-B_i C, B_i, C)$ corresponding to the situation when the first j loops are closed and the remaining are open. Let $W_C(S) = SI - A + B C$ denote the characteristic polynomial of S_0 . Now T_i maps D into T_i and t_s maps D into Y_s . Then

$$T_i(S) = W_C^i(S) / W_0(S)$$

and let N_i, n_i denote respectively the no. of ne encirclements of the origin by T_i and Y_i :

$$N_i = \sum_{j=1}^i n_j \quad (3.19)$$

This is from the theorem (2), which is

$$\text{For } i = 1 \dots m \\ T_i(S) = \prod_{j=1}^i t_j(S) \quad (3.20)$$

Now since $W_0(S)$ has no right half plane roots, the from eq.(3.19) N_i is equal to the number of roots of $W_C(S)$ in the right half plane. Thus the stability theorems are valid for scalar return difference and Scalar return ratios also. The proof (3.20) is in (6).

3.2. SEQUENTIAL RETURN DIFFERENCE ALGORITHM:

From the last discussion a design procedure, currently used in practice, would be to choose k_1 so that $t_1 = 1 + k_1 g_{11}$ is satisfactory, calculate G^1 , choose k_2 so that $t_2 = 1 + k_2 g_{22}$ is satisfactory. Calculate G^2 etc. However this procedure ignores the fact that even if $G_p(s)$ has no right half plane zeros, g_{ii} , $i=1\dots m$, as obtained above may right half plane zeros. The role of G_c , where $|G_c(s)|=1$ in the basic algorithm which has the objectives; performance and asymptotic stability. These zeros of right half plane give "design difficulty" appropriately to the various loops. The next is a simple sequential method for calculating G and t_i , $i=1\dots m$, which is SRD algorithm. The following algorithm generates G , t_i , $i=1\dots m$.

(i) Set $G(s) = G(s)$;

Set $i = 1$.

Choose $k_i(s)$

(ii) Set $t_i(s) = 1 + k_i(s) g_{ii}^{i-1}(s)$ (S)

(iii) If $i = n$, stop otherwise

Set $\bar{K}_i(s) = K_i(s) / t_i(s)$ (3.21)

Set $G(s) = G^{i-1}(s) - \bar{K}_i(s) g_{i-1}^{i-1}(s)$ (S) $g_{i-1}^{i-1}(s)$ (3.22)

Set $i = i + 1$

(iv) GO TO (ii)

Proof. For equation (3.22)

$$\begin{aligned} T_i(S) &= T_{i-1}(S) + K_i(S) g_i(S) g_{i.}(S) \\ &= T_{i-1}(S) \left[I_m + K_i(S) g_{i.}^{i-1}(S) g_{i.}^{i-1}(S) \right] \end{aligned}$$

Hence using a well known identity to invert the term in brackets (6).

$$T^{-i}(S) G(S) = \left[I_m - \overline{K}_i(S) g_{i.}^{i-1}(S) g_{i.}^{i-1}(S) \right] G^{i-1}(S) \quad (3.23)$$

$$G(S) = T^{-1} G(S)$$

$$\text{Thus } G(S) = G^{i-1}(S) - \overline{K}_i(S) g_{i.}^{i-1}(S) g_{i.}^{i-1}(S)$$

3.3. MODIFICATIONS MADE IN THE SRD ALGORITHM:

In the last step of sequential return difference algorithm $G(S)$ is calculated. $G(S)$ is the modified transfer function when first i loops are closed. If i is unity i.e. only one loop is closed, then only first diagonal element of transfer function $G(S)$ is going to be change. And if i is two, then only second diagonal element will change, and same will be the case for other values of i . So for reducing the computational complexities, only diagonal elements of the transfer function $G(S)$, which are modified are calculated. After this is done, another modification is that for closing each loop, n values of controller transfer function are assumed and corresponding scalar return differences and modified diagonal elements are calculated. If $i = 1 \dots n$, the scalar return difference is taken as $t(i, J)$, and in the same way the other variables. Thus the modified algorithm becomes:

The following algorithm generates $t_{i,j}^i(S)$, $G_{i,j}(S)$,
 $i=1 \dots n$, $J = 1 \dots n$

- (i) Set $G_{i-1,J}(S) = G(S)$
Set $i = 1$
- (ii) Choose $k_{i,j}(S) \quad J = 1 \dots n$
Set $t_{i,J}(S) = 1 + 1_{i,j}(S) \cdot g_{ij}(S), \quad J = 1 \dots n$
- (iii) Set $\bar{K}_{i,j}(S) = k_{i,j}(S) / t_{i,j}(S), \quad j=1 \dots n$
Set $g_{ii}^i(S) = g_{ii} - \bar{K}_{i,j}(S) g_i(S) g_i(S), \quad J = 1 \dots n$
Set $G_{i,j}(S) = G(S)$ with $g_{ii}(S)$ as $g_{ii}(S),$
 $j = 1 \dots n$
- (iv) If $i = m$ stop otherwise
Set $i = i + 1$
- (v) GO TO (ii)

For the development of computer program the elements of the transfer function $G(S)$ are named as $G_{11}^N(I), G_{11}^D(I), G_{12}^N(I), G_{12}^D(I) \dots \dots \dots$ etc. This is done because the elements of the transfer function are polynomials in S , so their coefficient arrays in descending order are taken for computations. Controller transfer functions are also polynomials in S , thus it becomes necessary to read them as $KN(L, I)$ and $KD(L, I)$, L stands for the number of values to be taken into account for computations, I is for the number of coefficient in descending order for a particular value of controller transfer function. N and D are for numerator and denominator.

CHAPTER -- IV

DESIGN OF REACTOR CONTROLLER

The state equations of the Nuclear reactor are given. For using sequential return difference algorithm, the given state equations are first converted into the matrix transfer function. Matrix transfer function is used in the algorithm described in 3.3. Thus the values of scalar return differences and the modified matrix transfer functions are achieved. Scalar return differences are used to check the stability. Out of the stable set of systems under consideration, few are checked for transient responses to step inputs.

4.1. TRANSFER FUNCTION OF THE NUCLEAR REACTOR :

The state equations for the nuclear reactor are given as :

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -0.188 & 0.0 & 0.227 \\ 0.0 & 0.0 & 1.0 \\ -2.138 & -0.587 & -0.550 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \\ u_3 \end{bmatrix}$$

State vectors, x_1, x_2 and x_3 are :

$x_1 = \text{E.m.f.}$

$x_2 = \text{rate of change of rotor angle}$

$x_3 = \text{rate of change of angular velocity}$

u_1, u_2 and u_3 are step inputs.

in the above representation

$$A = \begin{bmatrix} -0.188 & 0.0 & 0.227 \\ 0.0 & 0.0 & 1.0 \\ -2.138 & -0.587 & -0.550 \end{bmatrix} \quad B = \begin{bmatrix} 1.0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

Plant transfer function $G_p(s) = (sI - A)^{-1} B$

$$G(s) = G_p(s) G_c(s)$$

$$\text{Let } G_c(s) = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

Then, $G(s)$ is calculated as :

$$\begin{aligned} G_{11}N &= s^2 + 0.550s + 0.587, & G_{11}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{12}N &= -0.133 & G_{12}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{13}N &= 0.227s & G_{13}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{21}N &= 2.138 & G_{21}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{22}N &= s^2 + 0.738s + 0.5872 & G_{22}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{23}N &= s + 0.188 & G_{23}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{31}N &= 2.138 & G_{31}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{32}N &= -0.587(s + 0.188) & G_{32}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \\ G_{33}N &= s(s + 0.1888) & G_{33}D &= s^3 + 0.738s^2 + 1.1742s + 0.111 \end{aligned}$$

4.2. CONTROLLER TRANSFER FUNCTIONS ASSUMED AND RESULTING SYSTEMS:

For controller transfer functions, five values for each feed back loop have been assumed. The table is given below :

Table 4.1.

Loops	1	2	3	4	5
K_1	5/S	70	20	10	5
K_2	11	11	11	11	11
K_3	11	11	11	11	11

These values are given in the form $KN(L,I)$ and $KD(L,I)$, $I=1..2$, $L = 1...15$. The computer program for sequential return difference algorithm is given in Appendix A. Thus using above fifteen values we may arrive at many controllers, as three values are to be chosen for one controller. These controllers have to be checked for stability.

4.3. FINDING OUT THE STABILITY OF RESULTING SYSTEMS USING SCALAR RETURN DIFFERENCE:

From the results of computer program given in Appendix-A, the values of scalar return differences are found. These scalar return differences are used to check the stability by Routh Hurwitz criterion. The Sub-Routine for Routh's criterion is given in Appendix-A. For ensuring the tight feedback, a computer program is given in Appendix-B. In checking the feedback, again scalar return difference has been utilized which has been achieved from the results of sequential return difference algorithm.

Table 4.2

Sl.No.	A. WHEN FIRST FEEDBACK LOOP IS CLOSED		
	CONTROLLER TRANSFER FUNCTION	STABILITY	NATURE OF FEEDBACK
1.	5/S	UNSTABLE	TIGHT
2.	70	STABLE	TIGHT
3.	20	STABLE	TIGHT
4.	10	STABLE	TIGHT
5.	5	STABLE	TIGHT
Sl.No.	B. WHEN SECOND LOOP IS CLOSED		
	SAME	SAME	SAME
1 to 5	SAME	SAME	SAME
Sl.No.	C. WHEN THIRD LOOP IS CLOSED		
	SAME	SAME	SAME
1 to 5	SAME	SAME	SAME

From Table 4.2. it is clear that out of fifteen chosen values of controllers transfer functions, the system is stable only for twelve values. Thus we are getting four controller transfer functions for each feedback loop for a stable system.

4.4. PERFORMANCE

Having found out the values of controller transfer functions for which the system is stable, it becomes necessary to check which one gives good performance and with which feedback loop. Here, as in the present problem the order of the modified elements of modified transfer function matrix rises too much. Because of this rise in the order, it is very difficult to check the transient response of all the systems with modified transfer functions. So, here the Nyquist plot of all the systems are given. Thus relative stability is giving the measure of better performance. Program for Nyquist criterion is given in Appendix-C.

4.4.1. Step Response Calculations :

The values of controller transfer functions for each feedback loop giving best performance out of the chosen values of controller transfer functions for which system is stable, are $K(1,2)$, $K(2,2)$ and $K(3,3)$. The respective modified matrix transfer functions may be achieved from the results of the program given in Appendix-A. These becomes the data for the program of step response given in Appendix-D. As the result of step response, three polynomials of around thirty three order numerator and denominator are obtained. For comparison in transient

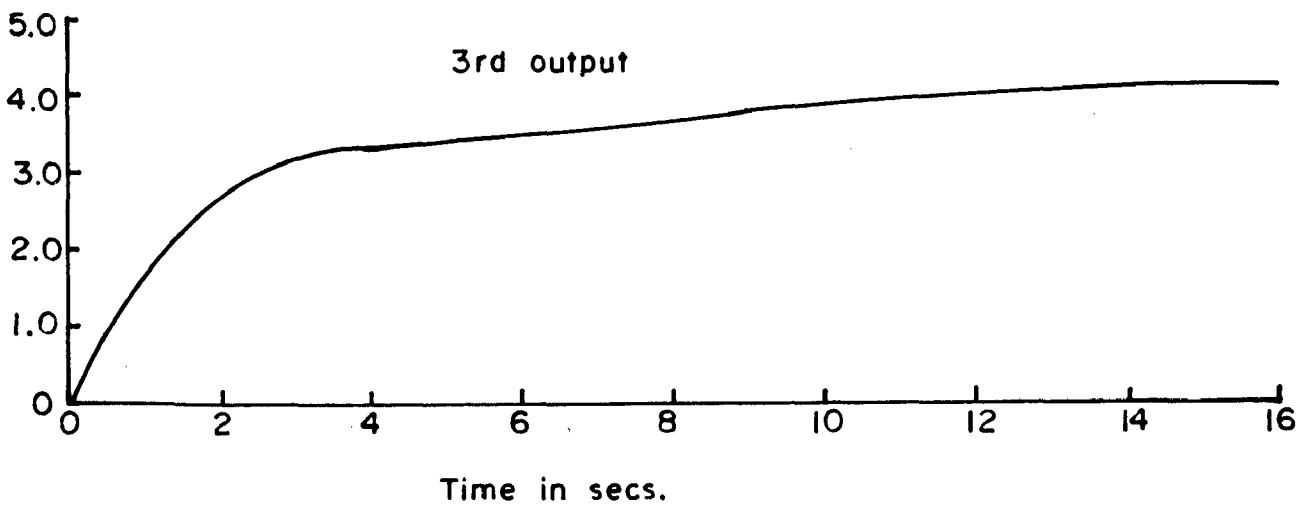
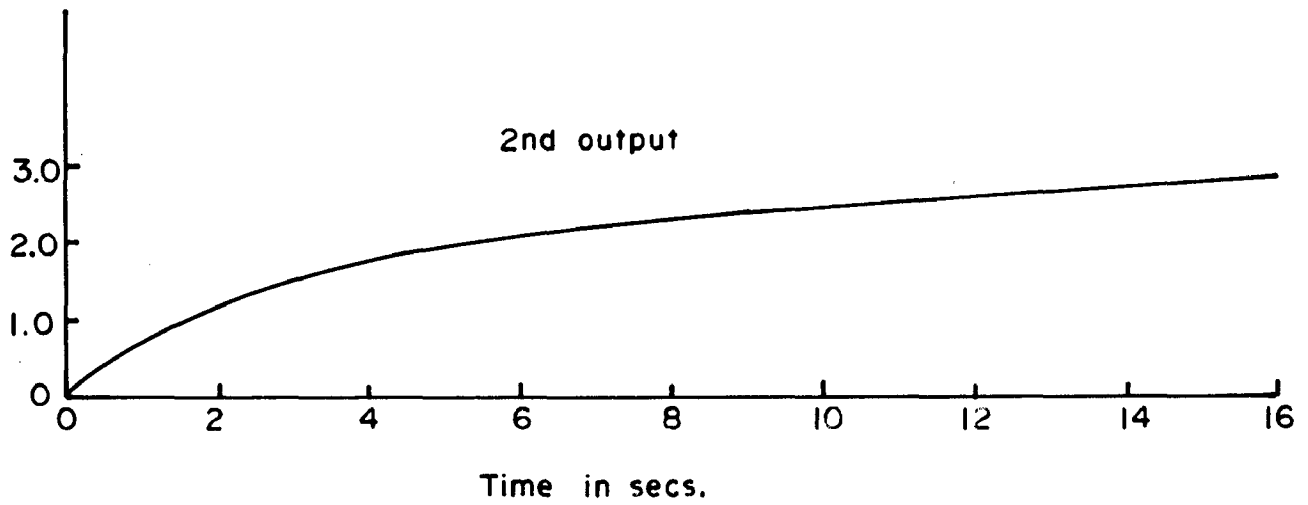
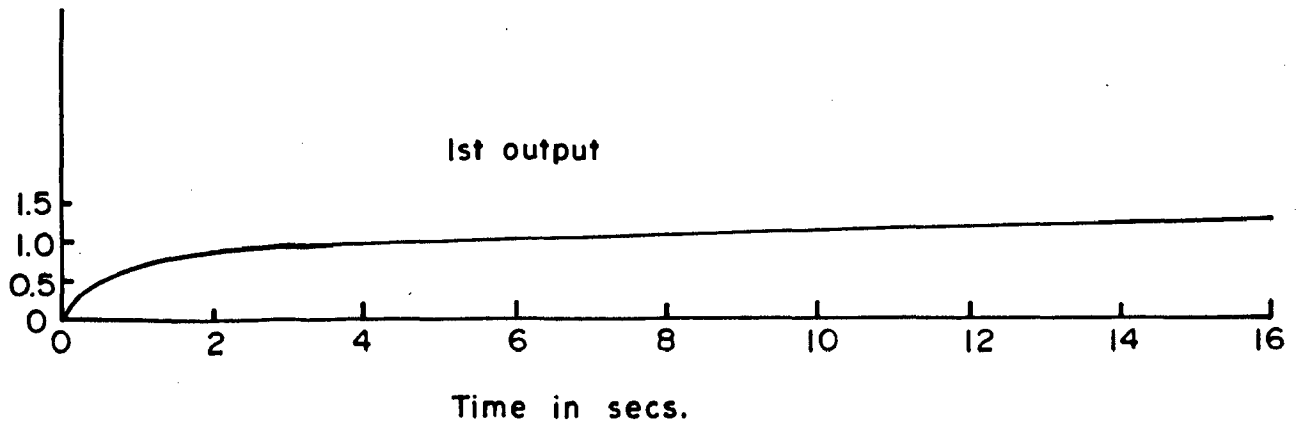


FIG. 5.7 TIME RESPONSES WITHOUT FEEDBACK.

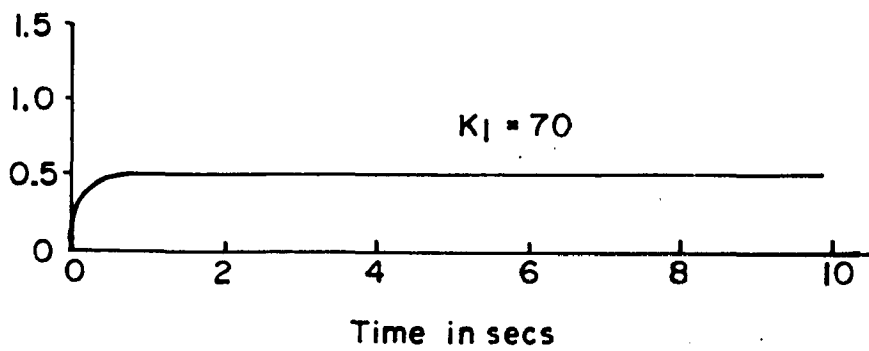
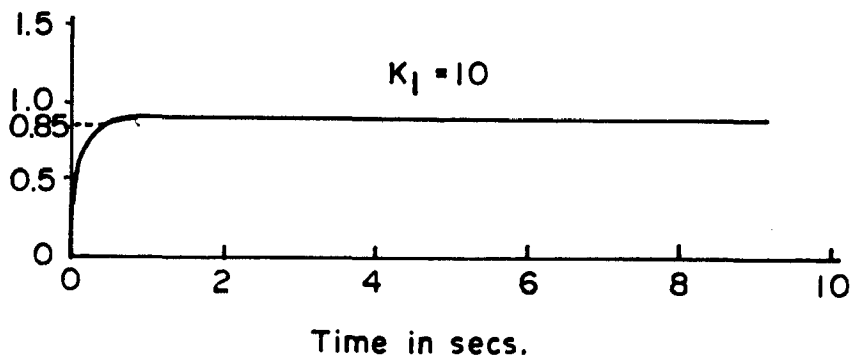
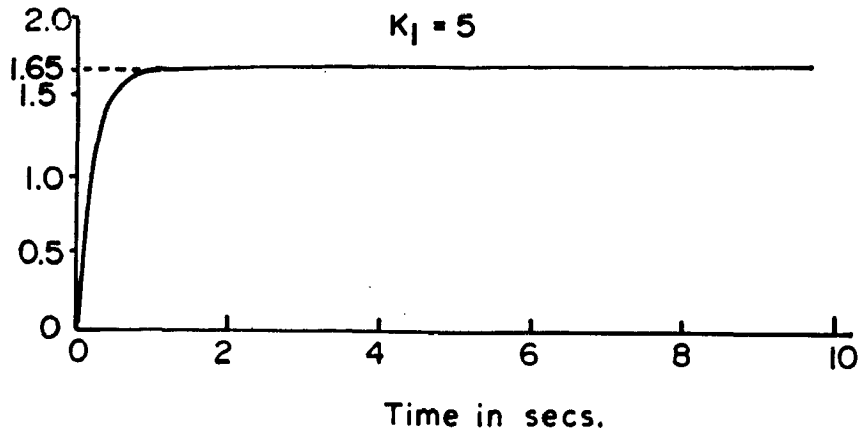


FIG.5.4 TIME RESPONSES FOR FIRST LOOP CLOSED.

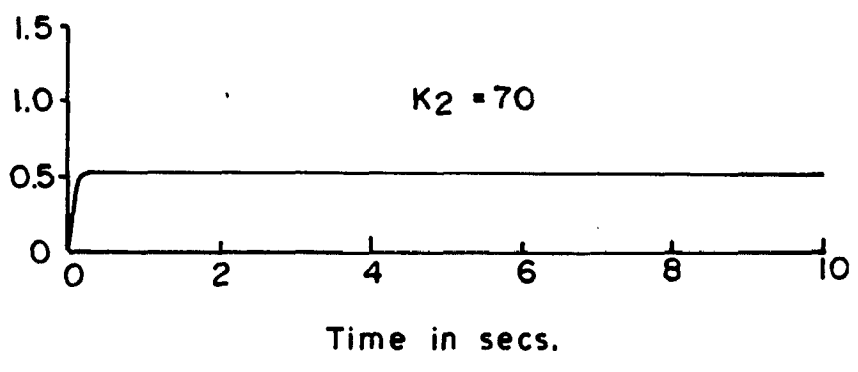
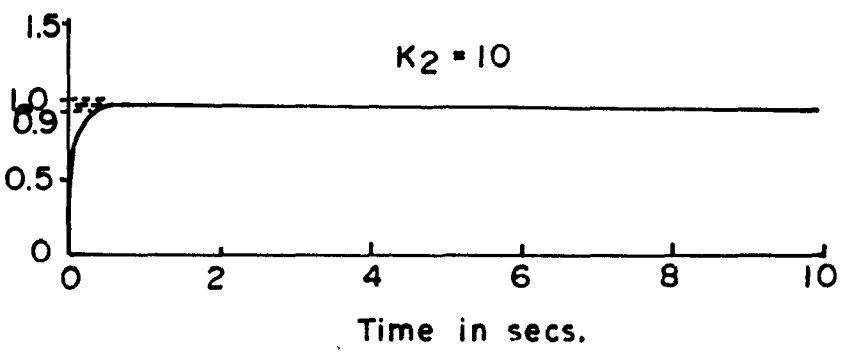
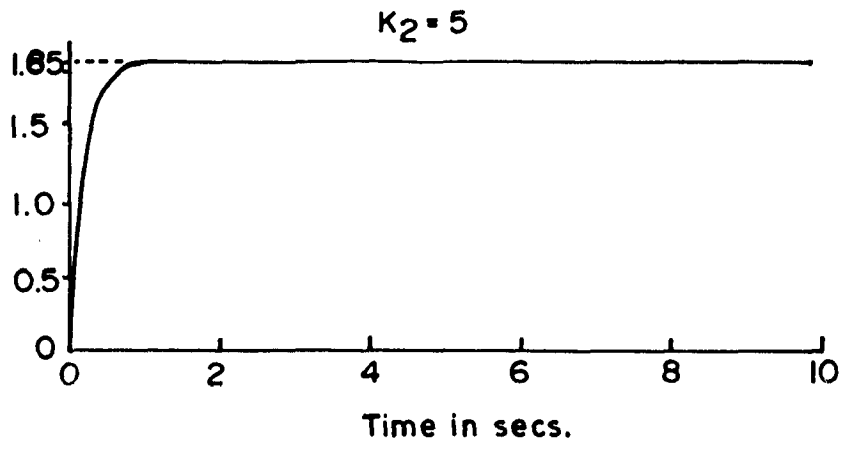


FIG.5.5 TIME RESPONSES FOR SECOND LOOP CLOSED.

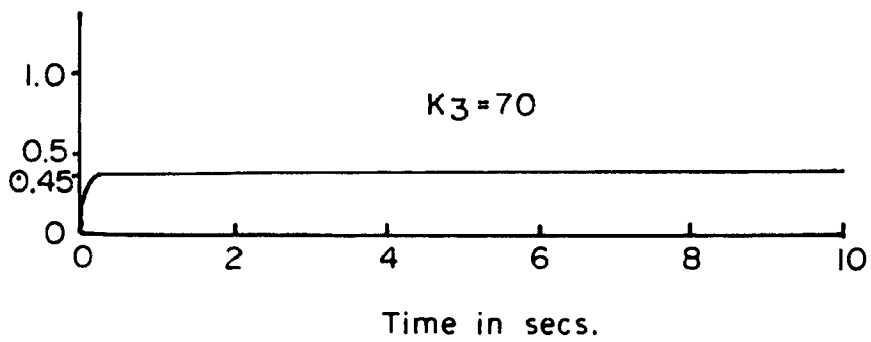
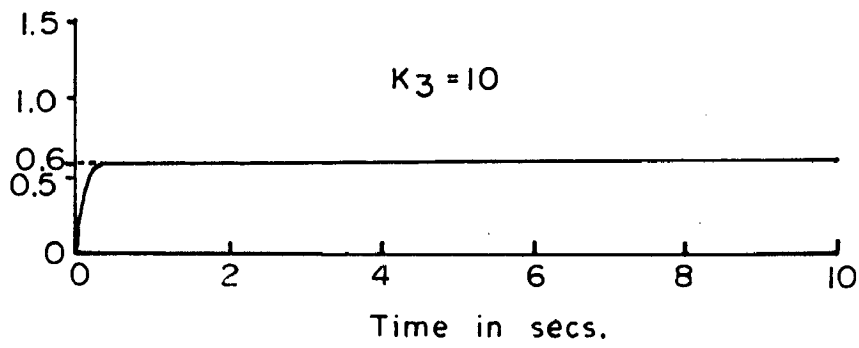
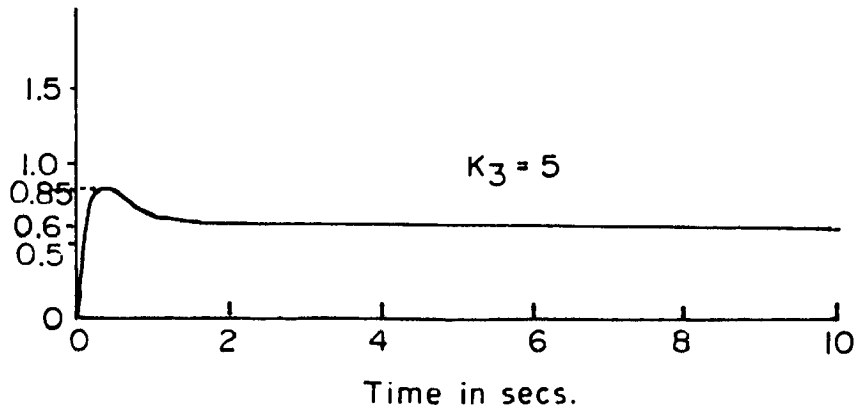


FIG.5.6 TIME RESPONSES FOR THIRD LOOP CLOSED.

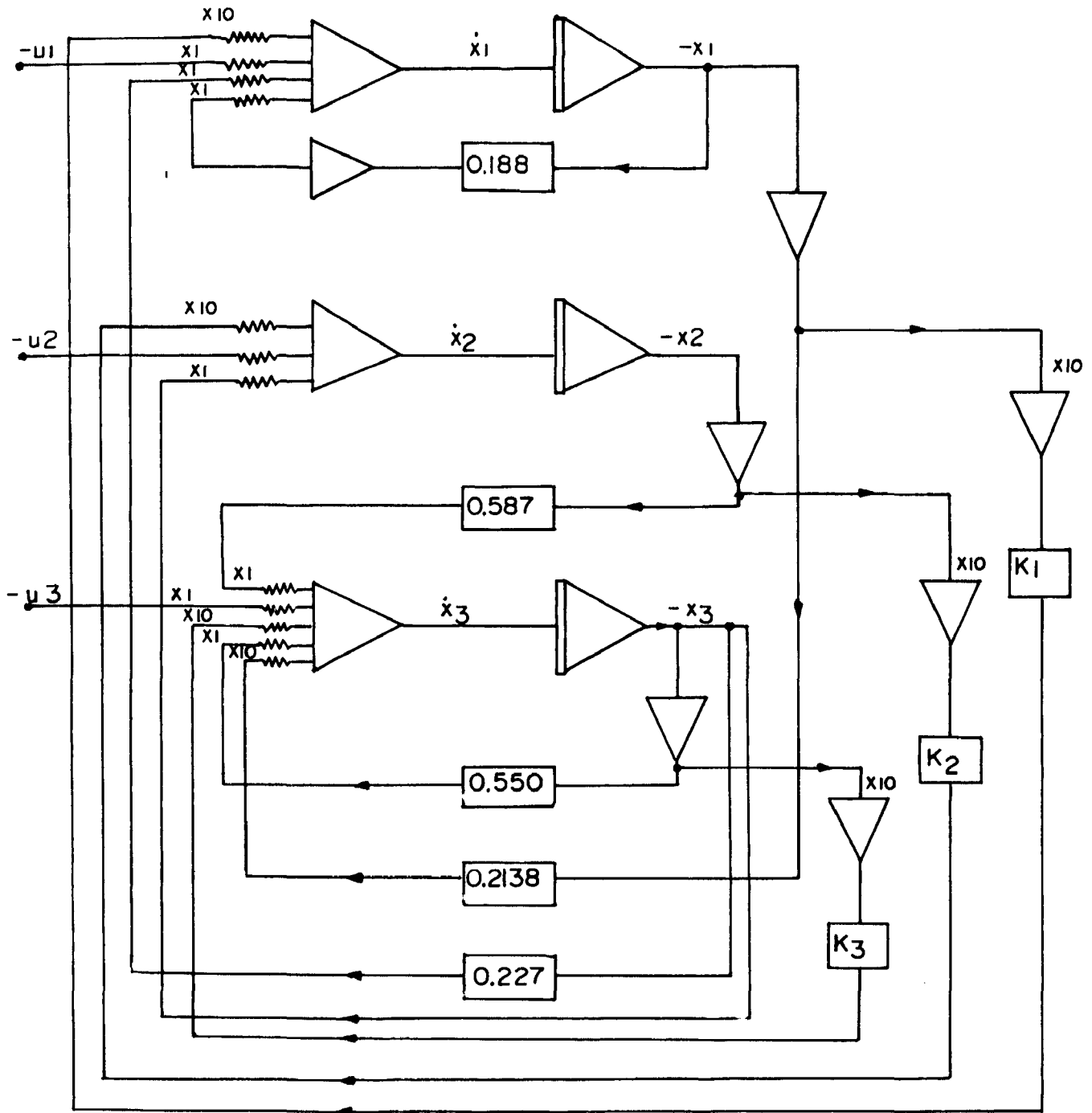


FIG.5.8

responses, the controller transfer functions $K(1,3)$, $K(2,3)$ and $K(3,3)$ are also considered for step responses.

4.4.2. Partial Fractions of the Step Response :

Before going to partial fractions, all the roots of the denominator of each response are determined. The program for roots is given in Appendix-E. These roots and the numerators of the responses are used in the program for partial fractions given in Appendix-F.

4.4.3. Transient Response :

The roots of the denominator of each response and their respective residues are the data for the program of "Laplace inverse and time response generation. This program is developed to give three sets of transient responses and each set giving two curves (for two values of controller gain) for one output. The open-loop response of the system with step-input is calculated and is shown in fig.5.7.

The closed-loop response of the plant is found for different controller settings, with the help of Analog Computer AC 20. The reason for using the Analog Computer is the unusual high time required by the digital computer in finding the roots of the high-order polynomials that are generated in the SED algorithm.

As seen from the closed-loop response from Figs.5.4,5.5,5.6, the response is very fast with no overshoot (compared with the sluggish response of the open-loop system). It must be, however, emphasised that the closed loop response has large steady-state error. This is

due to using very simple controller dynamics i.e. a single gain term only. With a single gain term in the feed back path such high values of steady-state errors are expected. To do away with this error in the response obviously one has to include further terms added in the feedback like integral terms and derivative terms. This was not tried in the present work only due to non-availability of a fast digital computer.

The closed-loop response was also found using the computer program developed. However, these results show high magnitudes (of the order of 10^3) for a step input. The reason is that on the application of SRD algorithm, the order of the modified diagonal elements rises upto 27. As a result some zeros of the modified diagonal elements happen to be in the right half plane and makes the element non-minimum phase. This is a disadvantage which is inherent with this algorithm.

These right half plane zeros of the modified diagonal elements may be cancelled by proper selection of compensating matrix $G_c(S)$. The value $G_c(S)$ should present case, the selection of $G_c(S)$ becomes difficult. Here, the order of modified diagonal elements is very high, and a number of zeros are in the right half plane. For cancellation of these zeros, the structure of compensating matrix will also be complex. A method to systematically determine $G_c(S)$ is given in 4.5.

GENERALIZED

4.5. A NEW GENERALIZED ALGORITHM FOR DESIGNING THE COMPENSATOR-
 $G_c(S)$:

For designing the compensator first one has to know the number of zeros coming in the right half plane, of the diagonal elements of modified transfer function matrix. In the present work the order of the numerator and denominator polynomials is about 27. IBM 1620 takes 25 to 40 minutes for finding out the real and imaginary roots of such a polynomial under the accuracy of 0.01. First consider the case of a three input three output system. If first diagonal element has the zeros A and B in right half plane, second diagonal element has the zeros C and D in right half plane and the third diagonal element has the zeros E and F in right half plane; then compensator has to be designed such that, it should have the poles at the locations of zeros in the respective diagonal element. The necessary condition is that the $G_c(S) = 1$. The possible compensator for the present problem what the author could think of is

$$G_c(S) = \begin{bmatrix} \frac{1}{(S-A)(S-B)} & 1 & 0 \\ 0 & \frac{1}{(S-C)(S-D)} & 1 \\ -1 \frac{1}{(S-C)(S-D)(S-E)(S-F)} & \frac{1}{(S-E)(S-F)} & \end{bmatrix}$$

$$\text{then } G_c(S) = 1$$

This is the case of three input three output system and with two zeros of each diagonal element in the right half plane. But in actual cases the problem may be some what different. Thus, there should be a generalized method for the design of compensator. A possible generalized matrix for the design of compensator is presented below :

Let $A_1(I)$ = No. of pole locations for first diagonal element of the compensator

$A_N(I)$ = No. of pole locations for Nth diagonal element

$$X(J) = \prod_{I=1}^{IZ(j)} (S - A(J, I))$$

Where

$$J = 1 \dots N$$

N is the number of inputs and outputs

$IZ(i)$ is the number of zeros in the right half plane for j -th diagonal element

Thus $IZ(1)$

$$X(1) = \prod_{I=1}^{IZ(1)} (S - A(1, I))$$

$$X(2) = \prod_{I=1}^{IZ(2)} (S - A(2, I))$$

$$X(N) = \prod_{I=1}^{IZ(N)} (S - A(N, I))$$

Now compensator may be given as

$$G_c(S) = \begin{bmatrix} \frac{1}{X(1)} & 1 & 0 & \dots\dots\dots 0 & 0 \\ 0 & \frac{1}{X(2)} & 1 & 0\dots\dots\dots 0 & 0 \\ 0 & 0 & \frac{1}{X(3)} & 1 & 0\dots\dots 0 & 0 \\ \vdots & \vdots & \vdots & & & \vdots \\ 0 & 0 & 0 & \frac{1}{X(N-1)} & & 1 \\ -1 & 0 & 0 & \frac{1}{X(N-1) \cdot X(N)} & & \frac{1}{X(N)} \end{bmatrix}$$

$$G_c(S) = 1$$

mathematically

$$G_c(j,j) = 1/X(j), \quad j= 1,N$$

$$G_c(j,j+1) = 1, \quad j= 1,N-1$$

$$G_c(N,1) = -1$$

$$G_c(N,N-1) = 1/X(N). X(N-1)$$

Thus, obtained compensating matrix is to be multiplied with modified transfer function matrix to provide phase minimality to the diagonal elements of the modified transfer function matrix.

But in the present case the order of the numerator and denominator is very high, and computer involved in the work being IBM 1620 which is very slow. Thus lot of computer time at this computer will be wasted only in checking the right half plane zeros. After that again finding the roots of the denominators of compensated diagonal elements and finding difficult to get the transient responses after compensation with this computer. The method is very useful after employing the algebraic operations for reducing the order of the polynomials which is at research level upto now. So performance is decided on the basis of relative stability by NYQUIST.

CHAPTER- V

RESULTS AND DISCUSSION

The results of sequential return difference algorithm and of other programs are given in the end. For the values of controller transfer functions with which the system is stable, Nyquist plots are given. Three sets of plots are for three output of the reactor, and each set is having four curves for different values of controller transfer functions. The best value of controller transfer function from each set of four curves is selected on the basis of relative stability.

These zeros of modified diagonal element coming in the right half plane introduce " design difficulty " which comes across only at the time of checking the transient response.

5.1. BRIEF INTRODUCTION TO COMPUTER PROGRAMS

First program is for sequential return difference method where it is made to read the original transfer function matrix element wise i.e. first it will read numerator of first element then denominator of the same and proceeds to the next element of first row, and same is for the remaining rows. Controller transfer functions are read in the same way. Subroutine of Routh Hurwitz criterion is called in the program to check the stability. Subroutine takes the return difference polynomial in the array of the coefficients of a polynomial in descending order. If some starting coefficients of the array are zero, then arrangement to

remove them and to reduce the order accordingly is made in the subroutine. After algebraic operations employment, the program will work with little modifications according to the operations.

Ensuring the tight feedback is second program where numerators and denominators of scalar return difference may be given in arrays A(I) and B (I). Arrangement for removal of initial zeros in the arrays is made here also.

Nyquist plot program also accepts the arrays as in the program of tight feedback. Here if some wants to get polar value i.e. r and Q, he can have by getting punched XM and XA.

For step response, first the modified diagonal elements will be supplied, then the remaining elements of the transfer function matrix. Subroutine PMULT is for polynomial multiplication and subroutine PSUM is for polynomial summation. This program gives in first iteration numerator, denominator for all the three outputs. And again in the second iteration gives all the things for second set. Note that this program reads matrix transfer function elements except the diagonal elements in another way. Such as for any element it will one coefficient from the numerator coefficients array and then one from denominator coefficients array and so on.

In the case of finding the roots, Bairstraw method has been employed. Here also you **give** the polynomial array in descending order.

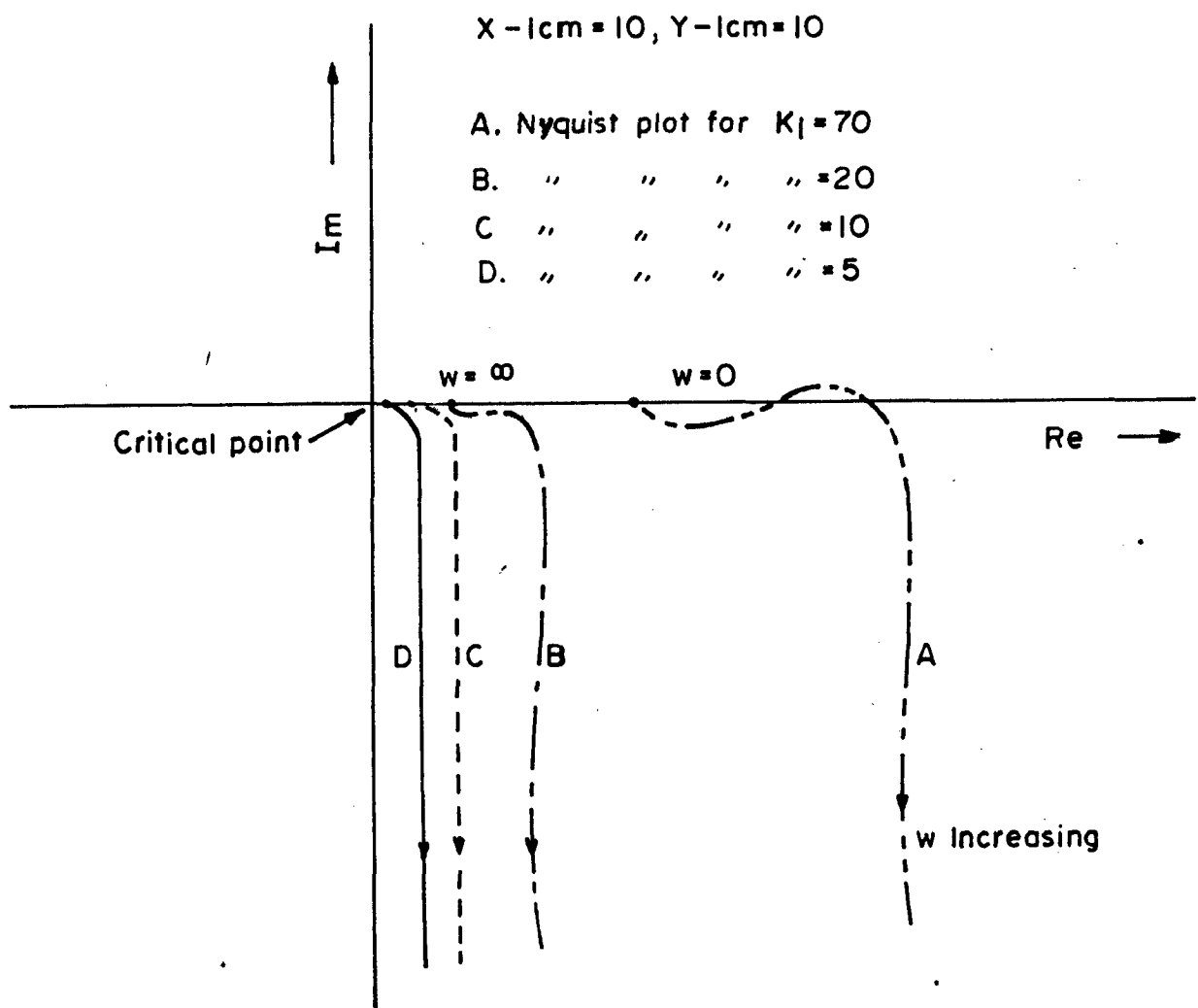


FIG.5.1 NYQUIST PLOTS FOR DIFFERENT VALUES OF CONTROLLER TRANSFER FUNCTIONS WHEN 1st LOOP IS CLOSED.

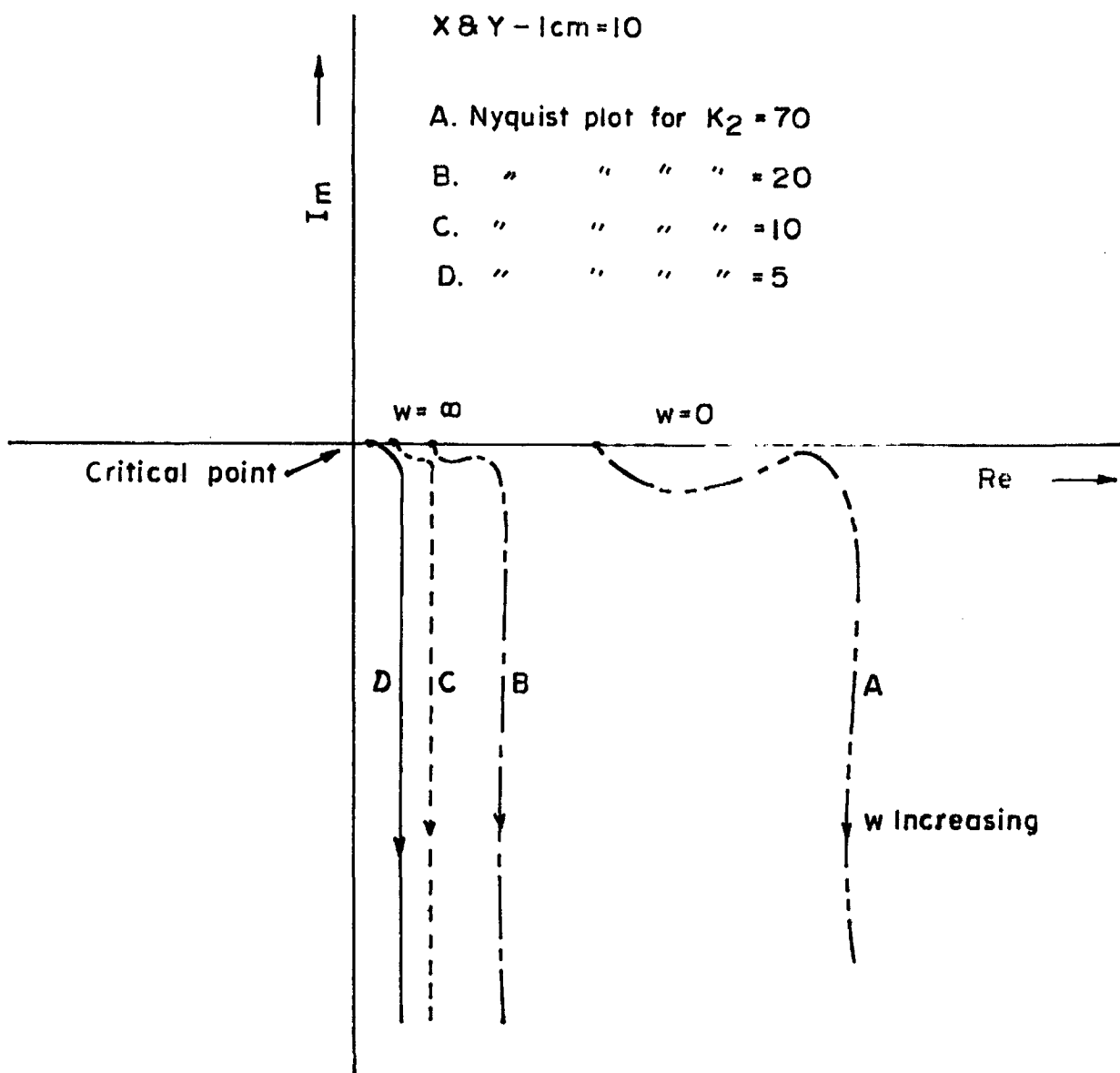


FIG.5.2 NYQUIST PLOT FOR DIFFERENT VALUES OF CONTROLLER TRANSFER FUNCTIONS WHEN 2nd LOOP IS CLOSED.

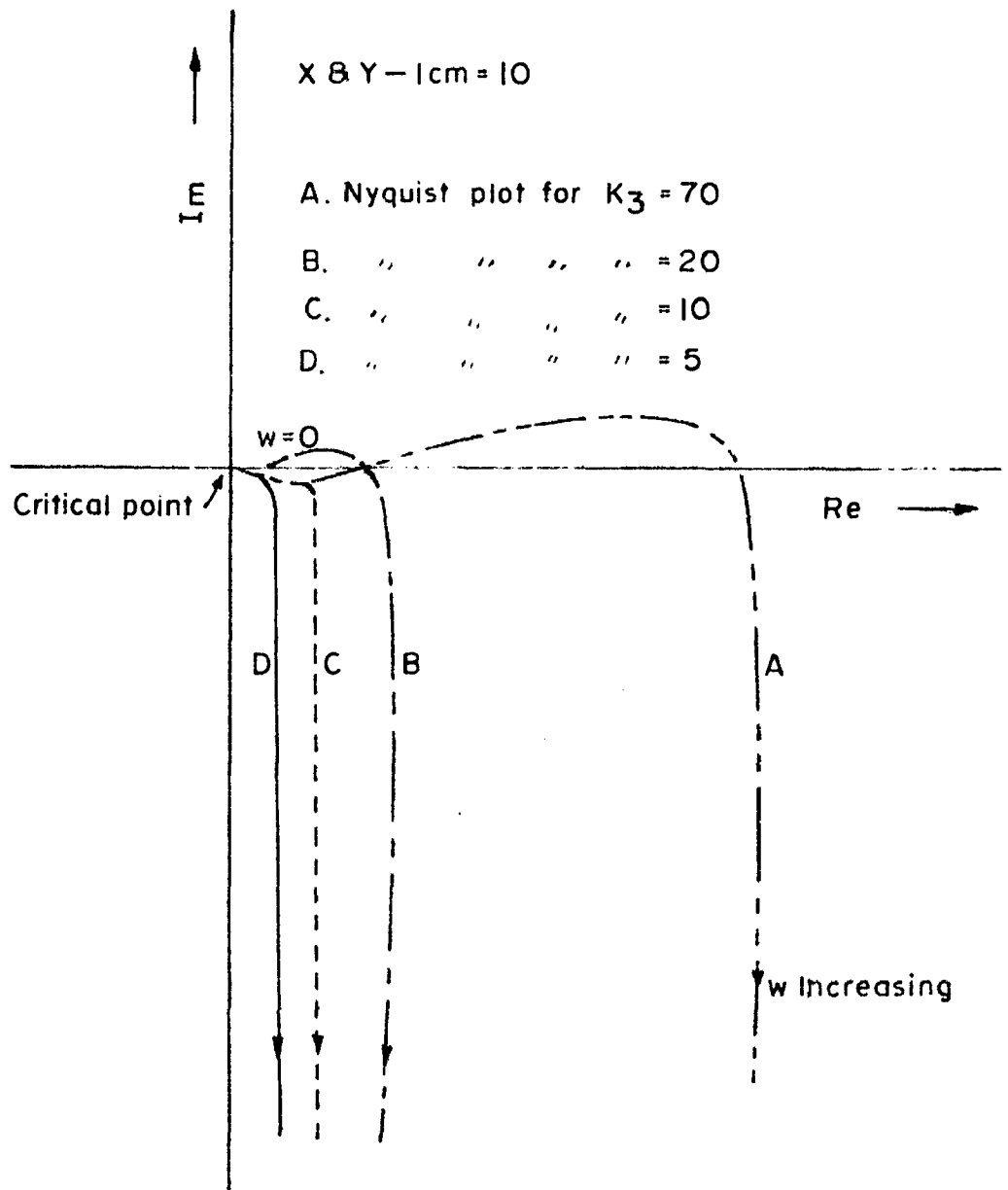


FIG.5.3 NYQUIST PLOTS FOR DIFFERENT VALUES OF CONTROLLER TRANSFER FUNCTIONS WHEN 3rd LOOP IS CLOSED.

Partial fractions may be obtained by providing the numerator polynomial array, array of real roots, array of imaginary roots and root multiplicity array. Here provision is made to prepare the data for the program of transients response. Thus in result alternate cards will be the data for the program of transient response.

Transient response program reads $A(I)$, $B(I)$, $C(I)$, $D(I)$ which are real part of root, real part of residue at this root, imaginary part of the root and imaginary part of the residue. $E(I)$, $F(I)$, $G(I)$, $H(I)$, are the same things for next polynomial of the same output. This program will give the transient response of the output with which the above two polynomials are associated. If you want transient responses for first output, give the value of $L = -1$, for second output $L = 0$, and for third output $L = 1$.

5.2. CONTROLLER PROPOSED FOR THE REACTOR:

On the basis of results of sequential return difference algorithm, Nyquist criterion and after ensuring the tight feedback, three values of controller transfer functions K_1 , K_2 and K_3 for all the three feedback loops are 70, 70 and 70 respectively. These controller transfer functions provide three tight feedback loops and a stable system for better performance out of the chosen controller transfer functions. The Nyquist plots for all the three controller transfer functions are given in fig. 5.1, 5.2 and 5.3 respectively.

5.3. SCOPE OF FURTHER WORK :

In the present work, the order of Polynomials of the diagonal elements in the modified transfer function matrix rises too much. Some work has also been done to reduce the order by algebraic operations (3). Further work is possible in many branches of this Classico-Modern technique, such as finding the optimal sequence of loop closure considering the probability of failure of all the loops. Considerations of input, output transducer failure may be included as a part of the present algorithm. Inclusion of checking the phase minimality of modified diagonal elements and the design of compensator accordingly may give the compact and self sufficient algorithm. Further, the algorithm may be extended to optimal design as a iterative method. After all the above inclusions, the method requires the use of very fast computers with interactive graphic displays. Further work needs to be done in implementing the algorithm given in 4.5. PID type of controller transfer functions may also be incorporated.

CONCLUSION

In the present problem, the sequential return difference method has been applied which gives rise to very high order of numerator and denominator polynomials of the diagonal elements of modified transfer function matrix. This high order has given computational difficulties. Because of the slow computer used, at this stage it was very difficult to develop one more program for compensator and then to achieve the time response. Actually after some algebraic operations for reducing the order of polynomial, this method utilizing the compensator also will be very much useful, and the computational time may be reduced tremendously. This method is also capable of including the cases of transducer failure, which will increase the flexibility of the design method.

Interactive graphic display systems, if available along with a moderately fast computer, would make the design procedure feasible for higher order industrial systems.

Reliability consideration may also be included. The feedback loop which is most probable to transducer failure should be closed in the last. However, much work needs to be done for finding the optimal sequence of loop closure.

Had the computational time been small, all the programs could have been used as the subroutines of the first program. This way the design procedure would have become very simple. One would have required only to change K_1 , K_2 and K_3 , and would have obtained the responses.

APPENDIX - A

S.K. SHARMA	SEQUENTIAL RETURN DIFF. ALGOR9THM	SRD 000
FIRST DATA CARD	CONTAINS THE ORDERS	SRD 000
AFTER FIRST NEXT NINE CARDS	ARE TO READ THE TRANSF. FUNCT. MAT.	SRD 000
ELEVENTH TO TWENTY FIRST DATA CARDS	ARE FOR CONTROLLOR K1.	SRD 000
TWENTY FIRST TO THIRTIETH DATA CARDS	ARE FOR CONTROLLOR K2.	SRD 000
THIRTY FIRST TO FORTIETH DATA CARDS	ARE FOR CONTROLLOR K3.	SRD 000
READ1,N,M,N1,M1		SRD 000
FORMAT(4I10)		SRD 000
M IS THE ORDER OF THE DENOMINAT. OF THE ELEMENT OF TRANS. FUNCT.	MV,)	001
N IS THE ORDER OF THE NUM. OF THE ELEMENT OF TRANS. FUNCT. MAT.		SRD 001
N1 IS THE ORDER OF THE NUM. OF THE CONTROLLOR		SRD 001
M1 IS THE ORDER OF THE DENOMINAT. OF THE CONTROLLOR		SRD 001
DIMENSION G11N(4),G11D(4),G12N(4),G13N(4),G13D(4),G12D(4),		SRD 001
1 G21N(4),G21D(4),G22N(4),G22D(4),G23N(4),G23D(4),G31N(4),G31D(4)		SRD 001
DIMENSION RD(27),GN(27),GD(27),SPN(27),SPD(27),ZN2(27)		SRD 001
DIMENSION ZD1(27),ANZN(27),ANZD(27),RN1(27),RN2(27),RN(27)		SRD 001
DIMENSION XN(27),XD(27),YN(27),YD(27),ZN(27),ZD(27),ZN1(27)		SRD 001
DIMENSION G32N(4),G32D(4),G33N(4),G33D(4)		SRD 001
DIMENSION KN(60,2),KD(60,2),SM(27)		
IF(N-M) 989,988,989		SRD 002
IR=M+1		SRD 002
GO TO 990		SRD 002
IR=N+1		SRD 002
READ2,(G11N(I),I=1,IR),(G11D(I),I=1,IR)		SRD 002
READ2,(G12N(I),I=1,IR),(G12D(I),I=1,IR)		SRD 002
READ2,(G13N(I),I=1,IR),(G13D(I),I=1,IR)		SRD 002
READ2,(G21N(I),I=1,IR),(G21D(I),I=1,IR)		SRD 002
READ2,(G22N(I),I=1,IR),(G22D(I),I=1,IR)		SRD 002
READ2,(G23N(I),I=1,IR),(G23D(I),I=1,IR)		SRD 003
READ2,(G31N(I),I=1,IR),(G31D(I),I=1,IR)		SRD 003
READ2,(G32N(I),I=1,IR),(G32D(I),I=1,IR)		SRD 003
READ2,(G33N(I),I=1,IR),(G33D(I),I=1,IR)		SRD 003
IF(N1-M1) 991,991,992		SRD 003
IT=M1+1		SRD 003
GO TO 994		SRD 003
IT=N1+1		SRD 003
DO 987 L=1,15		RHC 000
READ2,(KN(L,I),I=1,IT),(KD(L,I),I=1,IT)		SRD 003
FORMAT (8F10.3)		SRD 004
IXN=N+1		SRD 004
IXD=M+1		SRD 004
IYN=N1+1		SRD 004
IYD=M1+1		SRD 004
DO 110 L=1,15		SRD 068
IF(L-6) 112,113,113		SRD 069
DO 114 I=1,IXN		SRD 004
G11N(I)=G22N(I)		SRD 004
G11D(I)=G22D(I)		SRD 005
G31N(I)=G32N(I)		SRD 005
G31D(I)=G32D(I)		SRD 005
G13N(I)=G23N(I)		SRD 005

G13D(I)=G23D(I)	SRD 005
IF(L-11) 112,116,116	SRD 069
DO 117 I=1,IXN	SRD 005
G11N(I)=G33N(I)	SRD 005
G11D(I)=G33D(I)	SRD 005
G21N(I)=G23N(I)	SRD 006
G21D(I)=G23D(I)	SRD 006
G12N(I)=G32N(I)	SRD 006
G12D(I)=G32D(I)	SRD 006
CONTINUE	SRD 006
DO 15 I=1,IXN	SRD 006
XN(I)=G11N(I)	SRD 006
5 XD(I)=G11D(I)	SRD 006
DO 16 I=1,IYN	SRD 006
YN(I)=KN(L,I)	SRD 006
YD(I)=KD(L,I)	SRD 007
FINDING THE SCALAR RETURNED DIFFERENCE	SRD 007
IF (IXN*IYN) 4,4,5	SRD 007
IZN=0	SRD 007
GO TO 9	SRD 007
IZN =IXN+IYN	SRD 007
DO 6 I=1,IZN	SRD 007
ZN(I)=0.0	SRD 007
DO 8 I=1,IXN	SRD 007
DO 8 J=1,IYN	SRD 007
K=I+J-1	SRD 008
ZN(K)=XN(I)*YN(J)+ZN(K)	SRD 008
CONTINUE	SRD 008
IZN=IZN-2	SRD 008
IF(IXD*IYD) 10,10,11	SRD 008
IZD=0	SRD 008
GO TO 14	SRD 008
IZD=IXD+IYD	SRD 008
DO 12 I=1,IZD	SRD 008
ZD(I)=0.0	SRD 008
DO 13 I=1,IXD	SRD 009
DO 13 J=1,IYD	SRD 009
K=I+J-1	SRD 009
ZD(K)=XD(I)*YD(J)+ZD(K)	SRD 009
CONTINUE	SRD 009
IZD=IZD-2	SRD 009
MXN=IZN+1	SRD 009
MYD=IZD+1	SRD 009
ND=MXN	SRD 009
IF (MXN =MYD) 17,18,18	SRD 009
ND=MYD	SRD 010
IF (ND) 19,19,20	SRD 010
DO 21 I=1,ND	SRD 010
IF (I-MXN) 22,22,23	SRD 010
IF (I-MYD) 24,24,25	SRD 010
ZN1(I)=ZN(I)+ZD(I)	SRD 010
GO TO 21	SRD 010
ZN1(I)=ZD(I)	SRD 010
GO TO 21	SRD 010
ZN1(I)=ZN (I)	SRD 010

CONTINUE	SRD 011
IZN1=ND-1	SRD 011
MD=IZD+1	SRD 011
PUNCH 940	SRD 011
FORMAT(79 (1H*))	SRD 069
PUNCH 928	SRD 011
FORMAT (2X,62HORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND D	SRD 011
INOMINATOR)	SRD 011
PUNCH 150, IZN1, IZD	SRD 011
FORMAT (2I10)	SRD 011
PUNCH 929	SRD 012
FORMAT (2X,22 (1H*))	SRD 012
PUNCH 930	SRD 012
FORMAT (2X,63HSCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR K	SRD 012
RESPECTIVELY)	SRD 012
PUNCH 151, (ZN1(I), I=1,ND), (ZD(K),K=1,MD)	SRD 012
FORMAT (8F10.3)	SRD 012
PUNCH 931	SRD 012
FORMAT (2X,70 (1H*))	SRD 012
MSN=IZN1	SRD 069
DO 801 I=1, IZN1	SRD 069
SM(I)=ZN1(I)	SRD 069
CALL IROUTH(ZN1, IZN1)	SRD 012
PUNCH 827	SRD 069
FORMAT (79 (1H*))	SRD 069
IZN1=MSN	SRD 070
DO 802 I=1, IZN1	SRD 070
2 ZN1(I)=SM(I)	SRD 070
FINDING THE MODIFIED TRANSFER FUNCTION MATRIX	SRD 012
MXN=IYN	SRD 012
MYN=IZD+1	SRD 012
IF(MXN*MYN) 27,27,28	SRD 012
MZN=0	SRD 012
GO TO 31	SRD 012
MZN=MXN+MYN	SRD 012
DO 29 I=1, MZN	SRD 012
ZN1(I)=0.0	SRD 012
DO 30 I=1, MXN	SRD 012
DO 30 J=1, MYN	SRD 012
K=I+J-1	SRD 014
ZN1(K)=ZN1(K)+YN(I)*ZD(J)	SRD 014
CONTINUE	SRD 014
MZN=MZN-2	SRD 014
MXD=IYD	SRD 014
MYD =IZN +1	SRD 014
IF (MXD*MYD) 32,32,33	SRD 014
MZD=0	SRD 014
GO TO 36	SRD 014
MZD=MXD+MYD	SRD 014
DO 34 I=1, MZD	SRD 014
ZD1(I)=0.0	SRD 014
DO 35 I=1, MYD	SRD 014
DO 35 J=1, MXD	SRD 014
K=I+J-1	SRD 014
ZD1(K)=ZD1(K)+YD(J)*ZN(I)	SRD 014
CONTINUE	SRD 014

MZD =MZD-2	SRD 01
DO 39 I=1,IXN	SRD 01
YN(I)=G11N(I)	SRD 01
DO 48 I=1,IXD	SRD 01
YD(I)= G11D(I)	SRD 01
LYN=N+1	SRD 01
LYD=M+1	SRD 01
LIZN=2	SRD 01
LIZD=2	SRD 01
DO 157 I=1,LIZN	SRD 01
ANZN(I)=0.0	SRD 01
DO 158 I=1,LIZD	SRD 01
ANZD(I)=1.0	SRD 01
IS=1	SRD 01
IF(IXN*LYN)40,40,41	SRD 01
IZN1=0	SRD 01
GO TO 44	SRD 01
IZN1=IXN+LYN	SRD 01
DO 42 I=1,IZN1	SRD 01
ZN(I)=0.0	SRD 01
DO 43 I=1,IXN	SRD 01
DO 43 J=1,LYN	SRD 01
K=I+J-1	SRD 01
ZN(K)=XN(I)*YN(J)+ZN(K)	SRD 01
CONTINUE	SRD 01
IZN1=IZN1-2	SRD 01
IF (IXD*LYD) 45,45,46	SRD 01
IZD1=0	SRD 01
GO TO 51	SRD 01
IZD1=IXD+LYD	SRD 01
DO 49 I=1,IZD1	SRD 01
ZD(I)=0.0	SRD 01
DO 50 I=1,IXD	SRD 01
DO 50 J=1,LYD	SRD 01
K=I+J-1	SRD 01
ZD(K)=XD(I)*YD(J)+ZD(K)	SRD 01
CONTINUE	SRD 01
IZD1=IZD1-2	SRD 01
IZN1=IZN1+1	SRD 01
IZD1=IZD1+1	SRD 01
IF (LIZD*IZN1) 120,120,121	SRD 01
IRN1=0	SRD 01
GO TO 124	SRD 01
IRN1=LIZD+IZN1	SRD 02
DO 122 I=1,IRN1	SRD 02
RN1(I)=0.0	SRD 02
DO 123 I=1,LIZD	SRD 02
DO 123 J=1,IZN1	SRD 02
K=I+J-1	SRD 02
RN1(K)=RN1(K)+ANZD(I)*ZN(J)	SRD 02
CONTINUE	SRD 02
IRN1=IRN1-2	SRD 02
IF(LIZN*IZD1) 125,125,126	SRD 02
IRN2=0	SRD 02
GO TO 129	SRD 02
IRN2=LIZN+IZD1	SRD 02

	DO 127 I=1,IRN2	SRD 021
127	RN2(I)=0.0	SRD 021
	DO 128 I=1,LIZN	SRD 021
	DO 128 J=1,IZD1	SRD 021
	K=I+J-1	SRD 021
	RN2(K)=RN2(K)+ANZN(I)*ZD(J)	SRD 021
128	CONTINUE	SRD 021
	IRN2=IRN2-2	SRD 022
129	KXN=IRN1+1	SRD 022
	KYN=IRN2+1	SRD 022
	ND=KXN	SRD 022
	IF(KXN=KYN)130,131,131	SRD 022
130	ND=KYN	SRD 022
131	IF(ND)132,132,133	SRD 022
133	DO134 I=1,ND	SRD 022
	IF(I=KXN)135,135,136	SRD 022
135	IF(I=KYN)137,137,138	SRD 022
137	RN(I)=RN1(I)+RN2(I)	SRD 023
	GOTO134	SRD 023
136	RN(I)=RN2(I)	SRD 023
	GOTO134	SRD 023
138	RN(I)=RN1(I)	SRD 023
134	CONTINUE	SRD 023
132	IRN=ND-1	SRD 023
	IF(LIZD*IZD1)139,139,140	SRD 023
139	IRD=0	SRD 023
	GOTO143	SRD 023
140	IRD=LIZD+IZD1	SRD 024
	DO141 I=1,IRD	SRD 024
141	RD(I)=0.0	SRD 024
	DO142 I=1,LIZD	SRD 024
	DO142 J=1,IZD1	SRD 024
	K=I+J-1	SRD 024
	RD(K)=RD(K)+ANZD(I)*ZD(J)	SRD 024
142	CONTINUE	SRD 024
	IRD=IRD-2	SRD 024
143	IMN=IRN+1	SRD 024
	IMD=IRD+1	SRD 025
	DO52 I=1,IMN	SRD 025
52	ANZN(I)=RN(I)	SRD 025
	DO 53 I=1,IMD	SRD 025
53	ANZD(I)=RD(I)	SRD 025
	LIZN=IMN	SRD 025
	LIZD=IMD	SRD 025
	IS=IS+1	SRD 025
	IF(IS-2)55,56,55	SRD 025
56	DO57 I=1,IXN	SRD 025
57	XN(I)=G12N(I)	SRD 026
	DO61 I=1,IXD	SRD 026
61	XD(I)=G12D(I)	SRD 026
	DO 59 I=1,LYN	SRD 026
59	YN(I)=G21N(I)	SRD 026
	DO 60 I=1,LYD	SRD 026
60	YD(I)=G21D(I)	SRD 026
	GO TO 58	SRD 026
55	IF(IS-3)62,63,62	SRD 026

63	DO 64 I=1,IXN	SRD 026
64	XN(I)=G13N(I)	SRD 027
	DO 65 I=1,IXD	SRD 027
65	XD(I)=G13D(I)	SRD 027
	DO 66 I=1,LYN	SRD 027
66	YN(I)=G31N(I)	SRD 027
	DO 67 I=1,LYD	SRD 027
67	YD(I)=G31D(I)	SRD 027
	GO TO 58	SRD 027
62	DO 68 I=1,IMN	SRD 027
	GN(I)=ANZN(I)	SRD 027
68	XN(I)=GN(I)	SRD 028
	DO 69 I=1,IMD	SRD 028
	GD(I)=ANZD(I)	SRD 028
69	XD(I)=GD(I)	SRD 028
	MZN=MZN+1	SRD 028
	MZD=MZD+1	SRD 028
	DO 70 I=1,MZN	SRD 028
70	YN(I)=ZN1(I)	SRD 028
	DO 71 I=1,MZD	SRD 028
71	YD(I)=ZD1(I)	SRD 028
	IF(IMN*MZN) 72,72,73	SRD 029
72	ISZ=0	SRD 029
	GO TO 76	SRD 029
73	ISZ =IMN+MZN	SRD 029
	DO74I=1,ISZ	SRD 029
74	SPN(I)=0.0	SRD 029
	DO 75 I=1,IMN	SRD 029
	DO 75 J=1,MZN	SRD 029
	K=I+J-1	SRD 029
	SPN(K)=SPN(K)+XN(I)*YN(J)	SRD 029
75	CONTINUE	SRD 030
	ISZ=ISZ-2	SRD 030
76	IF(IMD*MZD) 77,77,78	SRD 030
77	ITZ=0	SRD 030
	GO TO 81	SRD 030
78	ITZ=IMD+MZD	SRD 030
	DO 79 I=1,ITZ	SRD 030
79	SPD(I)=0.0	SRD 030
	DO 80 I=1,IMD	SRD 030
	DO 80 J=1,MZD	SRD 030
	K=I+J-1	SRD 031
	SPD(K)=SPD(K)+XD(I)*YD(J)	SRD 031
80	CONTINUE	SRD 031
	ITZ=ITZ-2	SRD 031
81	DO82I=1,IXN	SRD 031
	XN(I)=G11N(I)	SRD 031
82	XD(I)=G11D(I)	SRD 031
	IZ3=ISZ+1	SRD 031
	IZ4=ITZ+1	SRD 031
	DO 84 I=1,IZ3	SRD 031
84	YN(I)=SPN(I)	SRD 032
	DO 85 I=1,IZ4	SRD 032
85	YD(I)=SPD(I)	SRD 032
	IF(IXN*IZ4) 86,86,87	SRD 032
86	IZN1=0	SRD 032
	GO TO 90	SRD 032

87	IZN1=IXN+IZ4	SRD 031
	DO 88 I=1,IZN1	SRD 032
88	ZN1(I)=0.0	SRD 032
	DO 89 I=1,IXN	SRD 032
	DO 89 J=1,IZ4	SRD 032
	K=I+J-1	SRD 032
	ZN1(K)=ZN1(K)+XN(I)*YD(J)	SRD 032
89	CONTINUE	SRD 032
	IZN1=IZN1-2	SRD 032
90	IF (IXN*IZ3) 91,91,92	SRD 032
91	IZN2=0	SRD 032
	GO TO 95	SRD 032
92	IZN2=IXN+IZ3	SRD 032
	DO 93 I=1,IZN2	SRD 032
93	ZN2(I)=0.0	SRD 034
	DO 94 I=1,IXN	SRD 034
	DO 94 J=1,IZ3	SRD 034
	K=I+J-1	SRD 034
	ZN2(K)=ZN2(K)+XD(I)*YN(J)	SRD 034
94	CONTINUE	SRD 034
	IZN2=IZN2-2	SRD 034
95	IX=IZN1+1	SRD 034
	IY=IZN2+1	SRD 034
	ND=IX	SRD 034
	IF(IX-IY) 96,97,97	SRD 035
96	ND=IY	SRD 035
97	IF (ND) 98,98,99	SRD 035
99	DO 100 I=1,ND	SRD 035
	IF(I=IX) 101,101,102	SRD 035
101	IF(I=IY) 103,103,104	SRD 035
103	ZN(I)=ZN1(I)-ZN2(I)	SRD 035
	GO TO 100	SRD 035
102	ZN(I)=-ZN2(I)	SRD 035
	GO TO 100	SRD 035
104	ZN(I)=ZN1(I)	SRD 036
100	CONTINUE	SRD 036
98	IZN=ND-1	SRD 036
	IF(IXN*IZ4) 105,105,106	SRD 036
105	IZD=0	SRD 036
	GO TO 109	SRD 036
106	IZD=IXN+IZ4	SRD 036
	DO 107 I=1,IZD	SRD 036
107	ZD(I)=0.0	SRD 036
	DO 108 I=1,IXN	SRD 036
	DO 108 J=1,IZ4	SRD 037
	K=I+J-1	SRD 037
	ZD(K)=ZD(K)+XD(I)*YD(J)	SRD 037
108	CONTINUE	SRD 037
	IZD=IZD-2	SRD 037
	IZD1=IZD+1	SRD 037
	IZN1=IZN+1	SRD 037
	PUNCH 943	SRD 037
943	FORMAT(2X,62HORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL	SRD 037
	1 ELEMENT)	SRD 071
109	PUNCH 152,IZN,IZD	SRD 037
152	FORMAT (2I10)	SRD 037
	PUNCH 942	SRD 037

942	FORMAT(22 (1H*));	SRD 03
	PUNCH 944	SRD 03
944	FORMAT(2X,61HNUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE	SRD 03
	1 CTIVELY)	SRD 07
	PUNCH153,(ZN(I),I=1,IZN1),(ZD(I),I=1,IZD1)	SRD 03
153	FORMAT (8F10.3)	SRD 03
110	CONTINUE	SRD 03
	PUNCH 941	SRD 03
941	FORMAT(79 (1H*));	SRD 03
	STOP	SRD 03
	END	SRD 03

**		SRD 00
	SUBROUTINE IRCUT(H(A,N1)	RHC 5
	DIMENSION A(27),X(27,27),B(27)	RHC C5
C	N1 IS THE ORDER OF THE POLYNOMIALS	RHC 00
	N=N1+1	RHC 00
	K=1	RHC C5
105	IF(A(K)) 101,102,101	RHC C5
102	N=N-1	RHC C5
	DO 103 I=1,N	RHC C5
103	A(I)=A(I+1)	RHC C5
	GO TO 105	RHC C5
101	DO 3 I=1,N	RHC C5
	B(I)=A(I)	RHC 00
	S=(-1)**N	RHC 00
	IF(S) 40,40,41	RHC 00
40	N2=(N+1)/2	RHC 00
	GO TO 44	RHC 00
41	N2=N/2	RHC 00
44	DO 4 I=1,N2	RHC 00
	JJ=1	RHC 00
	LL=2*I-1	RHC 00
4	X(JJ,I)=B(LL)	RHC 00
	T=(-1)**N	RHC 00
	IF(T) 42,42,43	RHC 00
42	N3=(N-1)/2	RHC 00
	GO TO 45	RHC 00
43	N3=N/2	RHC 00
45	DO 5 I=1,N3	RHC 00
	JJ=2	RHC 00
	MM=2*I	RHC 00
5	X(JJ,I)=B(MM)	RHC 00
	IF(X(2,1)) 91,92,91	RHC 00
92	GO TO 8	RHC 00
91	DO 6 JJ=3,N	RHC 00
	K=N3-JJ+2	RHC 00
	IF(K) 80,81,81	RHC 00
80	KK=K+2	RHC 00
	GO TO 90	RHC 00
81	KK=K+1	RHC 00
90	DO 70 I=KK,N3	RHC 00
70	X(JJ,I)=0.0	RHC 00
	IF(K) 51,50,49	RHC 00
50	K=K+1	RHC 00
49	DO 6 I=1,K	RHC 00

6	X(JJ,I)=X(JJ-2,I+1)-(X(JJ-2,1)*X(JJ-1,I+1))/X(JJ-1,1)	RHC 0038
51	DO 7 JJ=1,N	RHC 0039
	IF(X(JJ,1)) 8,7,7	RHC 0037
7	CONTINUE	RHC 0038
	GO TO 9	RHC 0039
8	PUNCH 13	RHC 0040
	GO TO 100	RHC 0041
9	DO 10 I=1,N	RHC 0042
	IF(X(I,1)) 8,11,10	RHC 0043
10	CONTINUE	RHC 0044
	PUNCH 15	RHC 0045
	GO TO 100	RHC 0046
11	PUNCH 13	RHC 0047
15	FORMAT (15X,6HSYSTEM,10X,2HIS,10X,6HSTABLE)	RHC 0048
13	FORMAT (15X,6HSYSTEM,10X,2HIS,10X,8HUNSTABLE)	RHC 0049
100	RETURN	RHC 0050
	END	RHC 0051

APPENDIX - B

```

*****
C C S.K.SHARMA PROGRAM FOR ENSURING TIGHT FEEDBACK TIF 001
C SUBROUTINE TIFB(A,B,N,M) TIF 001
C THIS SUBROUTINE ENSURES THE TIGHT FEEDBACK TIF 001
C A(I) IS THE NUMERATOR OF SCALAR RETURN DIFFERENCE TIF 001
C B(I) IS THE DINOMINATOR OF SCALAR RETURN DIFFERENCE TIF 001
C N-1 IS THE ORDER OF NUM. AND DINO. /F SCALAR RETUREN DIFFERENCE TIF 001
C M IS THE NUMBER OF SYSTEMS TO BE CHECKED TIF 001
  DIMENSION A(10),B(10) TIF 001
  READ 1,N,M TIF 001
  1 FORMAT (2I10) TIF 001
  NN=N $ KK=1 $ MN=N TIF 001
  DO 2 II=1,M TIF 001
  READ 3,(A(I),I=1,N),(B(I),I=1,N) TIF 001
  3 FORMAT (8F10.3) TIF 001
  4 IF(A(KK)) 7,5,7 TIF 001
  5 N=N-1 $ DO 6 I=1,N TIF 001
  6 A(I)=A(I+1) $ GO TO 4 TIF 001
  7 IF(B(KK)) 10,8,10 TIF 001
  8 NN=NN-1 $ DO 9 I=1,NN TIF 002
  9 B(I)=B(I+1) $ GO TO 7 TIF 002
 10 W=50. $ AR=0. $ AI=0. $ BR=0. $ BI=0. $ DO 11 J=1,N $ L=N-J TIF 002
  IF(A(J)) 25,11,25 TIF 002
 25 IF((-1)**L) 15,15,12 TIF 002
 12 IF((-1)**L/2) 13,13,14 TIF 002
 13 AR=AR-A(J)*(W**L) $ GO TO 11 TIF 002
 14 AR=AR+A(J)*(W**L) $ GO TO 11 TIF 002
 15 AI=AI+A(J)*(W**L) TIF 002
 11 CONTINUE $ TN=SQRTF(AR**2+AI**2) $ DO 16 J=1,NN $ L=NN-J TIF 002
  IF(B(J)) 26,16,26 TIF 003
 26 JK=(-1)**L $ JL=(-1)**L/2 $ IF(JK) 20,20,17 TIF 003
 17 IF(JL) 18,18,19 TIF 003
 18 BR=BR-B(J)*(W**L) $ GO TO 16 TIF 003
 19 BR=BR+B(J)*(W**L) $ GO TO 16 TIF 003
 20 BI=BI+B(J)*(W**L) TIF 003
 16 CONTINUE $ TD=SQRTF(BR**2+BI**2) $ X=TN/TD $ X=1.+X $ Y=1./X TIF 003
  N=MN $ NN=MN TIF 001
  IF(Y-1.) 21,21,22 TIF 003
 21 PUNCH 23,II $ GO TO 2 TIF 003
 22 PUNCH 24,II TIF 003
 23 FORMAT (16H THE FEEDBACK OF ,I3,16H SYSTEM IS TIGHT) TIF 003
 24 FORMAT (16H THE FEEDBACK OF ,I3,20H SYSTEM IS NOT TIGHT) TIF 004
  2 CONTINUE $ STOP $ END TIF 004
*****

```


APPENDIX - C

```

*****
**
C C S.K.SHARMA PROGRAM FOR NYQUIST PLOT GENERATION
  READ 1,N,M
  1 FORMAT (2I10)
  DIMENSION A(10),B(10),X(200),Y(200)
  KK=1 $ MN=N $ NN=N
  PUNCH33 $ PUNCH 34 $ PUNCH 35 $ PUNCH 33
  33 FORMAT (80(1H-))
  34 FORMAT (4(1H*,3X,2HX-,4X,1H*,3X,2HY*,4X))
  35 FORMAT (8(1H*,9HCORDINATE))
  DO 2 I=1,M
  READ 3,(A(I),I=1,N),(B(J),J=1,N)
  3 FORMAT (8F10,3)
  4 IF(A(KK)) 7,5,7
  5 N=N-1 $ DO 6 I=1,N
  6 A(I)=A(I+1) $ GO TO 4
  7 IF(B(KK)) 10,8,10
  8 NN=N-1 $ DO 9 I=1,NN
  9 B(I)=B(I+1) $ GO TO 7
  10 W=0. $ DO 27 IJ=1,20
  AR=0. $ AI=0. $ BR=0. $ BI=0. $ DO 11 J=1,N $ L=N-J
  JK=(-1)**L $ JL=(-1)**L/2 $ IF(A(J)) 25,11,25
  25 IF(JK) 15,15,12
  12 IF(JL) 13,13,14
  13 AR=AR-A(J)*(W**L) $ GO TO 11
  14 AR=AR+A(J)*(W**L) $ GO TO 11
  15 AI=AI+A(J)*(W**L)
  11 CONTINUE $ TN=SQRTF(AR**2+AI**2) $ DO 16 J=1,NN $ L=NN-J
  IF(B(J)) 26,16,26
  26 JK=(-1)**L $ JL=(-1)**L/2 $ IF(JK) 20,20,17
  17 IF(JL) 18,18,19
  18 BR=BR-B(J)*(W**L) $ GO TO 16
  19 BR=BR+B(J)*(W**L) $ GO TO 16
  20 BI=BI+B(J)*(W**L)
  16 CONTINUE $ TD=SQRTF(BR**2+BI**2) $ XM=TN/TD $ XA=ATANF(AI/AR)-ATAN
  1F(BI/BR) $ X(IJ)=XM*COSF(XA) $ Y(IJ)=XM*SINF(XA)
  W=W+.5
  27 CONTINUE $ PUNCH 28,(X(IJ),Y(IJ),IJ=1,20 )
  28 FORMAT (8(1H*,F9,3))
  N=MN $ NN=MN
  2 CONTINUE $ STOP $ END

```

APPENDIX - D

```

*****
**
C C S.K. SHARMA STEP RESPONSE OF REACTER WITH CONTROLLOR STR 049
C N IS THE ORDER OF THE NUMERATOR OF THE ELEMENT OF TRANSFER FUNCTIOVT*AT49 STR 049
C M IS THE ORDER OF THE DINO. OF THE ELEMENT OF TRANSFER FUNCTION MAX,*AT50
C N1 IS THE ORDER OF THE NUMERATOR OF THE STEP INPUT X,* T50
C M1 IS THE ORDER OF THE DINOMINATOR OF STEP INPUT X,* T50
C XN, XD, ARE THE NUMERATOR AND DINOMINATOR OF STEP INPUT RESPECTIVE X,* T50
C L IS THE NUMBER OF ITERATIONS X,* T50
C FIRST ITERATION IS FOR K(1,3),K(2,3), AND K(3,3) X,* T50
C SECOND ITERATION IS FOR K(1,4),K(2,4), AND K(3,4) X,* T50
C THIRD ITERATION IS FOR K(1,19),K(2,19), AND K(3,19) X,* T50
C FOURTH ITERATION IS FOR K(1,20),K(2,20), AND K(3,20) X,* T50
C X1, X2, AND X3, ARE THE OUTPUTS OF THE NUCLEAR REACTER X,* T50
READ 1,N,M,N1,M1,L X,* T51
1 FORMAT(5I10) X,* T51
DIMENSION G11N(4,40),G11D(4,40),G22N(4,40),G22D(4,40),G33N(4,40) X,* T51
DIMENSION G33D(4,40),G1N(40),G1D(40),RN1(40),RN2(40),RN3(40) X,* T51
DIMENSION RD1(40),RD2(40),RD3(40),ZN1(40),ZN2(40),ZD1(40),YN(40) X,* T51
DIMENSION YN1(40),YN2(40),ZN(40),ZD(40),G12N(40),G13N(40),G12D(40) X,* T51
DIMENSION G13D(40),G21N(40),G21D(40),G23N(40),G23D(40),G31N(40) X,* T51
DIMENSION G31D(40),G32N(40),G32D(40),XD(40),XN(40) X,* T51
IR=4 X,* T51
IXN=N+1 X,* T51
IXD=M+1 X,* T52
IYN=N1+1 52
IYD=M1+1 52
$EUD=,((RI((*(I*(GU)D(J,I),I=1,IXN),J=1,L) 52
READ2,((G22N(J,I),G22D(J,I),I=1,IXN),J=1,L) STR 052-
READ2,((G33N(J,I),G33D(J,I),I=1,IXN),J=1,L) STR 052
READ 2,(G12N(I),G12D(I),I=1,IR) STR 052
READ 2,(G13N(I),G13D(I),I=1,IR) STR 052
READ 2,(G21N(I),G21D(I),I=1,IR) STR 052
READ 2,(G23N(I),G23D(I),I=1,IR) STR 052
READ 2,(G31N(I),G31D(I),I=1,IR) STR 053
READ 2,(G32N(I),G32D(I),I=1,IR) STR 053
READ 2,(XN(I),XD(I),I=1,IYN) STR 053
2 FORMAT (8F10.3) STR 053
DO 101 J=1,L STR 053
DO 16 I=1,IXN STR 053
16 G1N(I)=G11N(J,I) STR 053
DO 17 I=1,IXD STR 053
17 G1D(I)=G11D(J,I) STR 053
IS=1 STR 053
100 CALL PMULT (RN1,IRN1,G1N,IXN,XN,IYN) STR 054
CALL PMULT (RD1,IRD1,G1D,IXD,XD,IYD) STR 054
CALL PMULT (RN2,IRN2,G12N,IR,XN,IYN) STR 054
CALL PMULT (RD2,IRD2,G12D,IR,XD,IYD) STR 054
CALL PMULT (RN3,IRN3,G13N,IR,XN,IYN) STR 054
CALL PMULT (RD3,IRD3,G13D,IR,XD,IYD) STR 054
IRN1=IRN1+1 STR 054
IRN2=IRN2+1 STR 054
IRN3=IRN3+1 STR 054
IRD1=IRD1+1 STR 054
IRD2=IRD2+1 STR 055
IRD3=IRD3+1 STR 055
CALL PMULT (ZN1,IZN1,RD1,IRD1,RN2,IRN2) STR 055
CALL PMULT (ZN2,IZN2,RN1,IRN1,RD2,IRD2) STR 055

```

	CALL PMULT (ZD1,IZD1,RD1,IRD1,RD2,IRD2)	STR 01
	IZN1=IZN1+1	STR 01
	IZN2=IZN2+1	STR 01
	IZD1=IZD1+1	STR 01
	CALL PSUM (YN,ISN,ZN1,IZN1,ZN2,IZN2)	STR 01
	CALL PMULT (YN1,IYN1,ZD1,IZD1,RN3,IRN3)	STR 01
	ISN=ISN+1	STR 01
	IYN1=IYN1+1	STR 01
	CALL PMULT (YN2,IYN2,YN,ISN,RD3,IRD3)	STR 01
	CALL PMULT (ZD,IZD,ZD1,IZD1,RD3,IRD3)	STR 01
	IYN2=IYN2+1	STR 01
	CALL PSUM (ZN,IZN,YN1,IYN1,YN2,IYN2)	STR 01
	IZN3=IZN+1	STR 01
	IZD3=IZD+1	STR 01
	ITRAN=J	STR 01
	PUNCH 70,ITRAN	STR 01
70	FORMAT (10X,9HITERATION,10X,6HNUMBER,10X,2HIS,5X,I4)	STR 01
	IF (IS-1) 20,21,20	STR 01
21	PUNCH 71,IZN	STR 01
	PUNCH 72,IZD	STR 01
	PUNCH 222	STR 01
	PUNCH22,(ZN(I),I=1,IZN3)	STR 01
	PUNCH 232	STR 01
	PUNCH23,(ZD(I),I=1,IZD3)	STR 01
71	FORMAT (10X,5HORDER,10X,2HOF,7HXN1(I)=,5X,I4)	STR 01
222	FORMAT (10X,4HNUM,5X,2HOF,5X,5HFIRST,5X,6HOUTPUT,5X,3HX1,5X,	STR 01
	1 2HIS,5X,5HGIVEN,5X,3HAS,5X)	STR 01
22	FORMAT (8F10.3)	STR 01
72	FORMAT (10X,5HORDER,10X,2HOF,10X,7HXD1(I)=,5X,I4)	STR 01
232	FORMAT (10X,5HDINO,5X,2HOF,5X,5HFIRST,5X,6HOUTPUT,5X,3HX1,5X,	STR 01
	1 2HIS,5X,5HGIVEN,5X,3HAS,4X)	STR 01
23	FORMAT (8F10.3)	STR 01
20	IF (IS-2) 32,25,24	STR 01
25	PUNCH 73,IZN	STR 01
	PUNCH 74,IZD	STR 01
	PUNCH 262	STR 01
	PUNCH26,(ZN(I),I=1,IZN3)	STR 01
	PUNCH 272	STR 01
	PUNCH27,(ZD(I),I=1,IZD3)	STR 01
73	FORMAT (10X,5HORDER,10X,2HOF,10X,7HXN2(I)=,5X,I4)	STR 01
262	FORMAT (10X,4HNUM,5X,2HOF,5X,6HSECOND,5X,6HOUTPUT,5X,3HX2,5X,	STR 01
	1 2HIS,5X,5HGIVEN,5X,3HAS,4X)	STR 01
26	FORMAT (8F10.3)	STR 01
74	FORMAT (10X,5HORDER,10X,2HOF,10X,7HXD2(I)=,5X,I4)	STR 01
272	FORMAT (10X,5HDINO,5X,2HOF,5X,6HSECOND,5X,6HOUTPUT,5X,3HX2,5X,	STR 01
	1 2HIS,5X,5HGIVEN,5X,3HAS,3X)	STR 01
27	FORMAT (8F10.3)	STR 01
24	IF (IS-3) 32,29,101	STR 01
29	PUNCH 75,IZN	STR 01
	PUNCH 76,IZD	STR 01
	PUNCH 302	STR 01
	PUNCH30,(ZN(I),I=1,IZN3)	STR 01
	PUNCH 312	STR 01
	PUNCH31,(ZD(I),I=1,IZD3)	STR 01
75	FORMAT (10X,5HORDER,10X,2HOF,10X,7HXN3(I)=,5X,I4)	STR 01

002	FORMAT (10X,4HNUM,5X,2HOF,5X,5HTHIRD,5X,6HOUTPUT,5X,3HX3,5X, 1 2HIS,5X,5HGIVEN,5X,3HAS,5X)	STR 060 STR 061
30	FORMAT (8F10,3)	STR 061
76	FORMAT (10X,5HORDER,10X,2HOF,10X,7HXD3(I)=,5X,I4)	STR 061
312	FORMAT (10X,5HDINO,5X,2HOF,5X,5HTHIRD,5X,6HOUTPUT,5X,3HX3,5X, 1 2HIS,5X,5HGIVEN,5X,3HAS,4X)	STR 061 STR 061
31	FORMAT (8F10,3)	STR 061
32	IS=IS+1	STR 061
	IF(IS-2)5,6,5	STR 061
6	DO 7 I=1,IXN	STR 061
7	G1N(I)=G22N(J,I)	STR 061
	DO 8 I=1,IXD	STR 062
8	G1D(I)=G22D(J,I)	STR 062
	DO 9 I=1,IR	STR 062
	G12N(I)=G21N(I)	STR 062
	G12D(I)=G21D(I)	STR 062
	G13N(I)=G23N(I)	STR 062
9	G13D(I)=G23D(I)	STR 062
	GO TO 100	STR 062
5	IF(IS-3)101,11,101	STR 062
11	DO 12 I=1,IXN	STR 062
12	G1N(I)=G33N(J,I)	STR 063
	DO 13 I=1,IXD	STR 063
13	G1D(I)=G33D(J,I)	STR 063
	DO 14 I=1,IR	STR 063
	G12N(I)=G31N(I)	STR 063
	G12D(I)=G31D(I)	STR 063
	G13N(I)=G32N(I)	STR 063
14	G13D(I)=G32D(I)	STR 063
	GO TO 100	STR 063
101	CONTINUE	STR 063
18	STOP	STR 064
	END	STR 064
*****		STR 064
**		STR 064
	SUBROUTINE PMULT (Z,IZ,C,IC,D,ID)	STR 064
C	THIS SUBROUTINE PERFORMS THE MULTIPLICATION OF TWO POLYNOMIALS	STR 064
	DIMENSION Z(40),C(40),D(40)	STR 064
	IX=IC	STR 064
	IY=ID	STR 064
	IF(IX*IY)50,50,51	STR 064
50	IZ=0	STR 064
	GO TO 54	STR 065
51	IZ=IX+IY	STR 065
	DO 52 I=1,IZ	STR 065
52	Z(I)=0.0	STR 065
	DO 53 I=1,IX	STR 065
	DO 53 J=1,IY	STR 065
	K=I+J-1	STR 065
	Z(K)=Z(K)+C(I)*D(J)	STR 065
53	CONTINUE	STR 065
	IZ=IZ-2	STR 065
54	RETURN	STR 066
	END	STR 066
*****		STR 066
**		STR 066

APPENDIX - E

```

*****
**
C C S.K.SHARMA FINDING THE ROOTS OF A POLYNOMIAL PRT 000
C N IS THE TOTAL NUMBER OF COEFFICIENTS IN THE POLYNOMIALS PRT 000
C P1 AND Q1 ARE THE STARTING VALUES OF THE ROOTS PRT 000
C E IS THE ACCURACY OF THE ROOTS PRT 000
  DIMENSION A(100),B(100),C(100) PRT 000
  READ 100,P1,Q1,E,N PRT 000
  READ 106,(A(I),I=1,N) PRT 000
  K=1
105 IF(A(K)) 101,102,101
102 N=N-1
  DO 103 I=1,N
103 A(I)=A(I+1)
  GO TO 105
101 N=N-1
  DO 12 I=1,N PRT 000
12 A(I)=A(I+1)/A(1) PRT 000
  LL=N PRT 000
  DO 17 K=1,N PRT 000
151 IF (A(LL)) 13,14,13 PRT 000
14 PUNCH 16 PRT 000
16 FORMAT (3H0NE,2X,4HROOT,2X,7H=0.0000) PRT 000
  LL=LL-1 PRT 000
17 CONTINUE PRT 000
13 LZ=K-1 PRT 000
  N=N-LZ PRT 000
  P=P1 PRT 000
  Q=Q1 PRT 000
  GO TO 75 PRT 000
10 B(1)=A(1)-P PRT 000
  B(2)=A(2)-P*B(1)-Q PRT 000
  DO 6 K=3,N PRT 000
6 B(K)=A(K)-P*B(K-1)-Q*B(K-2) PRT 000
  IF(N-3) 91,91,5 PRT 000
5 L=N-1 PRT 000
  C(1)=B(1)-P PRT 000
  C(2)=B(2)-P*C(1)-Q PRT 000
  DO 7 J=3,L PRT 000
7 C(J)=B(J)-P*C(J-1)-Q*C(J-2) PRT 000
  CBAR=C(L)-B(L) PRT 000
  DNR=(C(N-2))*2-CBAR*C(N-3) PRT 000
  IF(DNR) 15,500,15 PRT 000
500 P=P+1. PRT 000
  Q=Q+1. PRT 000
  GO TO 10 PRT 000
15 DLTP=(B(N-1)*C(N-2)-B(N)*C(N-3))/DNR PRT 000
9 DLTQ=(B(N)*C(N-2)-B(N-1)*CBAR)/DNR PRT 000
  P=P+DLTP PRT 000
  Q=Q+DLTQ PRT 000
  ABP=ABSF(DLTP) PRT 000
  ABQ=ABSF(DLTQ) PRT 000
  XM=ABP+ABQ PRT 000

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```

IF(XM-E) 50,50,10 PRT 00
C COMPUTING ROOTS OF QUADRATIC EQUATION PRT 00
50 DCRN=P*P-4.*Q PRT 00
IF(DCRN) 52,53,54 PRT 00
52 ABD=ABS(F(DCRN)) PRT 00
ABSQ=SQRT(ABD) PRT 00
XR=P/2. PRT 00
XIM=ABSQ/2. PRT 00
PUNCH 206,XR,XIM PRT 00
206 FORMAT(18HREAL PART OF ROOT=,F10.6,2X,15HIMAGINARY PART=,F10.6) PRT 00
GO TO 70 PRT 00
53 X=P/2. PRT 00
PUNCH 207,X PRT 00
207 FORMAT(12HEQUAL ROOTS=,F10.6) PRT 00
GO TO 70 PRT 00
54 ABSQ=SQRT(F(DCRN)) PRT 00
X1=-P/2.-ABSQ/2. PRT 00
X2=-P/2.+ABSQ/2. PRT 00
PUNCH 209,X1,X2 PRT 00
209 FORMAT(11HFIRST ROOT=,F10.6,12HSECOND ROOT=,F10.6) PRT 00
70 N=N-2 PRT 00
DO 71 K=1,N PRT 00
71 A(K)=B(K) PRT 00
75 IF(N)99,99,76 PRT 00
76 IF(N-4)77,77,10 PRT 00
77 GOTO (80,50,10,10),N PRT 00
80 X=-A(1) PRT 00
PUNCH 205,X PRT 00
GO TO 99 PRT 00
91 C(1)=B(1)-P PRT 00
C(2)=B(2)-P*C(1)-Q PRT 00
CBAR=C(L)-B(L) PRT 00
DNR=(C(N-2))*2-CBAR PRT 00
IF(DNR) 600,700,600 PRT 00
700 P=P+1. PRT 00
Q=Q+1. PRT 00
GO TO 91 PRT 00
600 DLTP=(B(N-1)*C(N-2)-B(N))/DNR PRT 00
GO TO 9 PRT 00
100 FORMAT(3F10.5,13) PRT 00
205 FORMAT(11HLAST ROOT=,F10.6) PRT 00
106 FORMAT(8F10.3) PRT 00
99 STOP PRT 00
END

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```
** PRF 00
C C S.K.SHARMA PARTIAL FRACTION EXPENSION PRF 00
C NNM IS THE ORDER OF THE NUMERATOR OF THE FUNCTION TO BE PARTIAL PRF 00
C FRACTIONED PRF 00
C IT IS THE NUMBER OF ROOTS WHICH ARE DISTINCT PRF 00
C NM IS THE TOTAL NUMBER OF FUNCTIONS TO BE PARTIAL FRACTIONED PRF 00
C MA IS THE ORDER OF THE NUMERATOR PLUS ONE PRF 00
C A(K,I) IS THE NUMERATOR POLYNOMIALS MATRIX PRF 00
C M(K,I) IS THE ROOT MULTIPLICITY MATRIX PRF 00
C CR(K,I) IS THE REAL ROOTS MATRIX OF THE POLYNOMIALS PRF 00
C CI(K,I) IS THE IMAGINARY ROOTS MATRIX OF THE POLYNOMIALS PRF 00
C RESR(K,I) IS THE MATRIX FOR THE REAL PART OF RESIDUES PRF 00
C RESI(K,I) IS THE MATRIX FOR THE IMAGINARY PARTS OF THE RESIDUES PRF 00
DIMENSION A(2,40),P(40),D(40),DX(40),M(2,40),Q(40),DXX(40) PRF 00
DIMENSION S2(40),ANS(40),AX(40),P1(40),P2(40),S1(40) PRF 00
DIMENSION RESR(40,8),RESI(40,8),CR(2,40),CI(2,40) PRF 00
READ1,NNM,IT,NM,MA PRF 00
1 FORMAT(4I10) PRF 00
DO 995 K=1,NM PRF 00
READ2,(A(K,I),I=1,MA) PRF 00
READ2,(CR(K,I),I=1,IT),(CI(K,I),I=1,IT) PRF 00
2 FORMAT(8F10.3) PRF 00
995 CONTINUE PRF 00
READ 222,((M(K,I),I=1,IT),K=1,NM) PRF 00
222 FORMAT (8I10) PRF 00
DO 1001 K=1,NM PRF 00
DO 199 I=1,MA PRF 00
D(I)=0.0 PRF 00
P(I)=0.0 PRF 00
Q(I)=0.0 PRF 00
DXX(I)=0.0 PRF 00
199 DX(I)=0.0 PRF 00
DO 4 L=1,MA PRF 00
4 AX(L)=A(K,L) PRF 00
KK=1 PRF 00
805 IF(AX(KK)) 801,802,801 PRF 00
802 MA=MA-1 PRF 00
DO 803 II=1,MA PRF 00
803 AX(II)=AX(II+1) PRF 00
GO TO 805 PRF 00
801 NNM=MA-1 PRF 00
200 DO 30 I=1,IT PRF 00
DO 5 L=1,MA PRF 00
P1(L)=AX(L) PRF 00
5 P2(L)=0.0 PRF 00
DO 6 L=1,IT PRF 00
S1(L)=CR(K,L) PRF 00
6 S2(L)=CI(K,L) PRF 00
C CALCULATING THE VALUE OF NUMERATOR POLYNOMIAL AT I,TH ROOT PRF 00
YR=0. PRF 00
YI=0. PRF 00
DO 7 J=1,MA PRF 00
L2=MA-J PRF 00
YR=YR+P1(J)*(S1(I)**L2) PRF 00
L3=(-1)**L2 PRF 00
IF(L3) 7003,7003,7000 PRF 00
7000 L4=(-1)**(L3/2) PRF 00
IF(L4) 7001,7001,7002 PRF 00
7001 YR=YR-P1(J)*(S1(I)**L2) PRF 00
GO TO 7 PRF 00
7002 YR=YR+P1(J)*(S1(I)**L2) PRF 00
GO TO 7 PRF 00
```


	IF(IQ1*ID1)10,10,20	PRF 011
10	IDX=0	PRF 011
	GO TO 50	PRF 011
20	IDX=IQ1+ID1	PRF 012
	DO 35 I=1,IDX	PRF 012
35	DX(I)=0.0	PRF 012
	DO 36 I=1,IQ1	PRF 012
	DO 36 J=1,ID1	PRF 012
	L=I+J-1	PRF 012
	DX(L)=DX(L)+Q(I)*D(J)	PRF 012
36	CONTINUE	PRF 012
	IDX=IDX-2	PRF 012
50	D(1)=CI(K,JJ)*CI(K,JJ)+CR(K,JJ)*CR(K,JJ)	PRF 012
	D(2)=2.0*CR(K,JJ)	PRF 013
	D(3)=1.0	PRF 013
	ID=2	PRF 013
	IP1=IP+1	PRF 013
	ID1=ID+1	PRF 013
	IF(IP1*ID1)51,51,52	PRF 013
51	IDXX=0	PRF 013
	GO TO 55	PRF 013
52	IDXX=IP1+ID1	PRF 013
	DO 53 I=1,IDXX	PRF 013
53	DXX(I)=0.0	PRF 014
	DO 54 I=1,IP1	PRF 014
	DO 54 J=1,ID1	PRF 014
	L=I+J-1	PRF 014
	DXX(L)=DXX(L)+P(I)*D(J)	PRF 014
54	CONTINUE	PRF 014
	IDXX=IDXX-2	PRF 014
55	IS=IDX+1	PRF 014
	IDX=IDXX+1	PRF 014
	ND=IS	PRF 014
	IF(IS-IDX) 56,57,57	PRF 015
56	ND=IDX	PRF 015
57	IF(ND)58,58,59	PRF 015
59	DO 60 I=1,ND	PRF 015
	IF(I-IS) 61,61,62	PRF 015
61	IF(I-IDX)63,63,64	PRF 015
63	P(I)=DX(I)+DXX(I)	PRF 015
	GO TO 60	PRF 015
62	P(I)=DXX(I)	PRF 015
	GO TO 60	PRF 015
64	P(I)=DX(I)	PRF 016
60	CONTINUE	PRF 016
58	IP=ND-1	PRF 016
	IR=IQ+1	PRF 016
	IY=ID+1	PRF 016
	IF(IR*IY)65,65,66	PRF 016
65	IX=0	PRF 016
	GO TO 69	PRF 016
66	IX=IR+IY	PRF 016
	DO 67 I=1,IX	PRF 016
67	DX(I)=0.0	PRF 017
	DO 68 I=1,IR	PRF 017
	DO 68 J=1,IY	PRF 017

7003	YI=YI+P1(J)*(S2(I)**L2)	PRF 00
7	CONTINUE	PRF 00
C	FINDING THE RESIDUES AT DISTINCT ROOTS	PRF 00
	PI=0.	PRF 00
	IMI=M(K,I)	PRF 00
	PR=1.	PRF 00
	IF(IMI)30,30,390	PRF 00
390	DO 31 J=1,IT	PRF 00
	IF(I=J)32,31,32	PRF 00
32	IMJ=M(K,J)	PRF 00
	IF(IMJ)31,31,33	PRF 00
33	A1=CR(K,I)-CR(K,J)	PRF 00
	B1=CI(K,I)-CI(K,J)	PRF 00
	DO 34 L=1,IMJ	PRF 00
	A2=PR*A1-PI*B1	PRF 00
	B2=PR*B1+PI*A1	PRF 00
	PR=A2	PRF 00
	PI=B2	PRF 00
34	CONTINUE	PRF 00
31	CONTINUE	PRF 00
	DIV=PR*PR+PI*PI	PRF 00
	RESR(I,IMI)=(PR*YR+PI*YI)/DIV	PRF 00
	RESI(I,IMI)=(PR*YI-PI*YR)/DIV	PRF 00
	PUNCH 528,RESR(I,IMI),RESI(I,IMI)	PRF 00
528	FORMAT (4HREAL,1X,4HPART,1X,2HOF,1X,9HRESIDUES=,E14.8,	PRF 00
	1 9HIMAGINARY,1X,4HPART,1X,2HOF,1X,9HRESIDUES=,E14.8)	PRF 00
	SK=S1(I)	PRF 00
	SS=S2(I)	PRF 00
	SKA=RESR(I,IMI)	PRF 00
	SA=RESI(I,IMI)	PRF 00
	PUNCH 7005,SK,SKA,SS,SA	PRF 00
7005	FORMAT(4F15.4)	PRF 00
30	CONTINUE	PRF 00
C	FINDING OUT THE RESIDUES FOR THE ROOT MULTIPLICITY	PRF 00
	DO 300 I=1,IT	PRF 00
	IF(M(K,I)-1)300,300,301	PRF 00
300	CONTINUE	PRF 00
	GO TO 205	PRF 00
301	JJ=0	PRF 01
	DO 40 I=1,MA	PRF 01
	Q(I)=0.0	PRF 01
	P(I)=0.0	PRF 01
40	CONTINUE	PRF 01
	IQ=0	PRF 01
	IP=0	PRF 01
	Q(1)=1.0	PRF 01
42	JJ=JJ+1	PRF 01
	ICH=M(K,JJ)	PRF 01
	IF(ICH)42,42,422	PRF 01
422	IF(CI(K,JJ))43,45,43	PRF 01
43	D(1)=-2.*RESI(JJ,ICH)*CI(K,JJ)-2.*RESR(JJ,ICH)*CR(K,JJ)	PRF 01
	D(2)=2.*RESR(JJ,ICH)	PRF 01
	ID=1	PRF 01
	IQ1=IQ+1	PRF 01
	ID1=ID+1	PRF 01

	IF(IQ1*ID1)10,10,20	PRF 011
10	IDX=0	PRF 011
	GO TO 50	PRF 011
20	IDX=IQ1+ID1	PRF 012
	DO 35 I=1,IDX	PRF 012
35	DX(I)=0.0	PRF 012
	DO 36 I=1,IQ1	PRF 012
	DO 36 J=1,IDI	PRF 012
	L=I+J-1	PRF 012
	DX(L)=DX(L)+Q(I)*D(J)	PRF 012
36	CONTINUE	PRF 012
	IDX=IDX-2	PRF 012
50	D(1)=CI(K,JJ)*CI(K,JJ)+CR(K,JJ)*CR(K,JJ)	PRF 012
	D(2)=2.*CR(K,JJ)	PRF 013
	D(3)=1.	PRF 013
	ID=2	PRF 013
	IP1=IP+1	PRF 013
	ID1=ID+1	PRF 013
	IF(IP1*ID1)51,51,52	PRF 013
51	IDXX=0	PRF 013
	GO TO 55	PRF 013
52	IDXX=IP1+ID1	PRF 013
	DO 53 I=1,IDXX	PRF 013
53	DXX(I)=0.0	PRF 014
	DO 54 I=1,IP1	PRF 014
	DO 54 J=1,IDI	PRF 014
	L=I+J-1	PRF 014
	DXX(L)=DXX(L)+P(I)*D(J)	PRF 014
54	CONTINUE	PRF 014
	IDXX=IDXX-2	PRF 014
55	IS=IDX+1	PRF 014
	IDX=IDXX+1	PRF 014
	ND=IS	PRF 014
	IF(IS-IDX) 56,57,57	PRF 015
56	ND=IDX	PRF 015
57	IF(ND)58,58,59	PRF 015
59	DO 60 I=1,ND	PRF 015
	IF(I-IS) 61,61,62	PRF 015
61	IF(I-IDX)63,63,64	PRF 015
63	P(I)=DX(I)+DXX(I)	PRF 015
	GO TO 60	PRF 015
62	P(I)=DXX(I)	PRF 015
	GO TO 60	PRF 015
64	P(I)=DX(I)	PRF 016
60	CONTINUE	PRF 016
58	IP=ND-1	PRF 016
	IR=IQ+1	PRF 016
	IY=ID+1	PRF 016
	IF(IR*IY)65,65,66	PRF 016
65	IX=0	PRF 016
	GO TO 69	PRF 016
66	IX=IR+IY	PRF 016
	DO 67 I=1,IX	PRF 016
67	DX(I)=0.0	PRF 017
	DO 68 I=1,IR	PRF 017
	DO 68 J=1,IY	PRF 017

	L=I+J-1	PRF 017
	DX(L)=DX(L)+Q(I)*D(J)	PRF 017
68	CONTINUE	PRF 017
	IX=IX-2	PRF 017
69	LL=IX+1	PRF 017
	DO 70 I=1,LL	PRF 017
70	Q(I)=DX(I)	PRF 017
	IQ=IX	PRF 018
	JJ=JJ+1	PRF 018
	IF(IT-JJ)71,71,42	PRF 018
45	D(1)=-CR(K,JJ)	PRF 018
	D(2)=1.	PRF 018
	ID=1	PRF 018
	DO 74 IN=1,ICH	PRF 018
	ID1=ID+1	PRF 018
	IP1=IP+1	PRF 018
	IF(ID1*ID1)75,75,76	PRF 018
75	IXX=0	PRF 019
	GO TO 79	PRF 019
76	IXX=0	PRF 019
761	IXX=ID1+IP1	PRF 019
	DO 77 I=1,IXX	PRF 019
77	DX(I)=0.	PRF 019
	DO 78 I=1,ID1	PRF 019
	DO 78 J=1,IP1	PRF 019
	L=I+J-1	PRF 019
	DX(L)=DX(L)+D(I)*P(J)	PRF 019
78	CONTINUE	PRF 020
	IXX=IXX-2	PRF 020
79	LL=IXX+1	PRF 020
	DO 81 I=1,LL	PRF 020
81	P(I)=DX(I)	PRF 020
	IP=IXX	PRF 020
74	CONTINUE	PRF 020
	IQP1=IQ+1	PRF 020
	DO 82 IXX=1,IQP1	PRF 020
	DXX(IXX)=RESR(JJ,ICH)*Q(IXX)	PRF 020
82	CONTINUE	PRF 021
	IK=IP	PRF 021
	IK1=IK+1	PRF 021
	IQ1=IQ+1	PRF 021
	ND=IK1	PRF 021
	IF(IK1-IQ1)83,84,84	PRF 021
83	ND=IQ1	PRF 021
84	IF(ND)85,85,86	PRF 021
86	DO 91 I=1,ND	PRF 021
	IF(I-IK1)87,87,88	PRF 021
87	IF(I-IQ1)89,89,90	PRF 022
89	P(I)=DX(I)+DXX(I)	PRF 022
	GO TO 91	PRF 022
88	P(I)=DXX(I)	PRF 022
	GO TO 91	PRF 022
90	P(I)=DX(I)	PRF 022
91	CONTINUE	PRF 022
85	IP=ND-1	PRF 022
	DO 49 IXX=1,ICH	PRF 022
	IX=ID+1	PRF 022

	IY=IQ+1	PRF 023
	IF(IX*IY)92,92,93	PRF 023
92	IKX=0	PRF 023
	GO TO 96	PRF 023
93	IKX=IX+IY	PRF 023
	DO 94 I=1,IKX	PRF 023
94	DX(I)=0.	PRF 023
	DO 95 I=1,IX	PRF 023
	DO 95 J=1,IY	PRF 023
	L=I+J-1	PRF 023
	DX(L)=DX(L)+D(I)*Q(J)	PRF 024
95	CONTINUE	PRF 024
	IKX=IKX-2	PRF 024
96	LL=IKX+1	PRF 024
	DO 97 I=1,LL	PRF 024
97	Q(I)=DX(I)	PRF 024
	IQ=IKX	PRF 024
49	CONTINUE	PRF 024
71	IX=NNM+1	PRF 024
	IY=IP+1	PRF 024
	ND=IX	PRF 025
	IF(IX-IY)98,99,99	PRF 025
98	ND=IY	PRF 025
99	IF(ND)100,100,101	PRF 025
101	DO 102 I=1,ND	PRF 025
	IF(I-IX)103,103,104	PRF 025
103	IF(I-IY)105,105,106	PRF 025
105	ANS(I)=AX(I)-P(I)	PRF 025
	GO TO 102	PRF 025
104	ANS(I)=-P(I)	PRF 025
	GO TO 102	PRF 026
106	ANS(I)=AX(I)	PRF 026
102	CONTINUE	PRF 026
100	IANS=ND-1	PRF 026
	LL=IANS+1	PRF 026
	DO 107 I=1,LL	PRF 026
107	AX(I)=ANS(I)	PRF 026
	NNM=IANS	PRF 026
	TOL=0.001	PRF 026
	I=0	PRF 026
501	I=I+1	PRF 027
	IF(I-IT) 701,701,200	PRF 027
701	IF(M(K,I)) 501,501,502	PRF 027
502	M(K,I)=M(K,I)-1	PRF 027
	IF(CI(K,I))507,504,507	PRF 027
507	D(1)=CR(K,I)*CR(K,I)+CI(K,I)*CI(K,I)	PRF 027
	D(2)=-2.*CR(K,I)	PRF 027
	D(3)=1.	PRF 027
	ID=2	PRF 027
	I=I+1	PRF 027
	IF(I-IT) 702,702,505	PRF 023
702	M(K,I)=M(K,I)-1	PRF 023
	GO TO 505	PRF 028
504	D(1)=-CR(K,I)	PRF 023
	D(2)=1.	PRF 023

	ID=1	PRF 02
505	IX=NNM+1	PRF 02
	IY=ID+1	PRF 02
508	IF(IY) 509,509,510	PRF 02
510	IF(ABS(D(IY))-TOL) 511,511,509	PRF 02
511	IY=IY-1	PRF 02
	GO TO 508	PRF 02
509	IF(IY) 512,512,513	PRF 02
513	IANS=IX-IY+1	PRF 02
	IF(IANS) 514,515,516	PRF 02
514	IANS=0	PRF 02
515	IER=0	PRF 02
517	GO TO 525	PRF 02
512	IER=1	PRF 02
	GO TO 525	PRF 02
516	IX=IY-1	PRF 02
	IA=IANS	PRF 02
519	II=IA+IX	PRF 02
	ANS(IA)=AX(II)/D(IY)	PRF 02
	IF(IX) 990,990,991	PRF 02
991	DO 520 L=1,IX	PRF 02
	LL=L-1+IA	PRF 02
	AX(LL)=AX(LL)-ANS(IA)*D(L)	PRF 02
520	CONTINUE	PRF 02
990	IA=IA-1	PRF 02
	IF(IA) 521,521,519	PRF 02
521	IF(IX) 522,522,523	PRF 02
523	IF(ABS(AX(IX))-TOL) 524,524,522	PRF 02
524	IX=IX-1	PRF 02
	GO TO 521	PRF 02
522	IANS=IANS-1	PRF 02
	GO TO 515	PRF 02
525	LL=IANS+1	PRF 02
	DO 526 IB=1,LL	PRF 02
526	AX(IB)=ANS(IB)	PRF 02
	NNM=IANS	PRF 02
	IF(I-IT) 501,200,200	PRF 02
1001	CONTINUE	PRF 02
205	STOP	PRF 02
	END	PRF 02

APPENDIX - G

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*****
**
C C S.K.SHARMA TRANSIENT RESPONSE CALCULATIONS AND PLOTTING
  DIMENSION XINE(120),X13(120),X14(120)
  DIMENSION A(40),B(40),C(40),D(40),E(40),F(40),G(40),H(40)
  READ 1,BLANK,DOT,X,STAR
1  FORMAT (4A1)
  READ 2,TMAX,XMAX,N,M,L,KL
2  FORMAT(2F10.3,4I10)
97  IF(L=2) 96,120,120
**
  DO 4006 JK=1,KL
  READ 4002,A(JK),B(JK),C(JK),D(JK)
4002 FORMAT(4F15.4)
4006 CONTINUE
  DO 4007 JK=1,KL
  READ 4004,E(JK),F(JK),G(JK),H(JK)
  4004 FORMAT(4F15.4)
4007 CONTINUE
96  IF(L) 229,230,231
229  PUNCH 232
232  FORMAT (10X,9HTRANSIENT,1X,9HRESPONSES,1X,3HFOR,1X,5HFIRST,1X,6
  1 HOUTPUT)
  GO TO 95
230  PUNCH 233
233  FORMAT (10X,9HTRANSIENT,1X,9HRESPONSES,1X,3HFOR,1X,6HSECOND,1X,6
  1HOUTPUT)
  GO TO 95
231  PUNCH 234
234  FORMAT (10X,9HTRANSIENT,1X,9HRESPONSES,1X,3HFOR,1X,5HTHIRD,1X,6
  1HOUTPUT)
  IF(TMAX) 95,120,95
95  JMAX=2*N+1
  DO 100 J=1,JMAX
100  XINE(J)=DOT
  PUNCH 3,(XINE(J),J=1,JMAX)
3  FORMAT (1H*,79A1)
  DO 105 J=1,JMAX
105  XINE(J)=BLANK
  T=0.0
  RM=M
  RN=N
  DELT=TMAX/RM
  DO 110 K=1,M
  T=T+DELT
  IF(L=2) 101,120,120
101  IF(L) 201,301,401
301  XMAX=2.*XMAX
  GO TO 201
401  XMAX=3.*XMAX
201  X13R=0.
  X13I=0.
  DO 4001 JK=1,KL
  RX=EXPF(A(JK)*T)*(B(JK)*COSF(C(JK)*T)-D(JK)*SINF(C(JK)*T))
  X13R=X13R+RX
  RY=EXPF(A(JK)*T)*(D(JK)*COSF(C(JK)*T)+B(JK)*SINF(C(JK)*T))
4001 X13I=X13I+RY

```

	X13(K)=SQRTF(X13R**2+X13I**2)	TRS	00
	X14R=0.	TRS	00
	X14I=0.	TRS	00
	DO 4003 JK=1,KL	TRS	00
	RZ=EXPF(E(JK)*T)*(F(JK)*COSF(G(JK)*T)-H(JK)*SINF(G(JK)*T))	TRS	00
	X14R=X14R+RZ	TRS	00
	SZ=EXPF(E(JK)*T)*(H(JK)*COSF(G(JK)*T)+F(JK)*SINF(G(JK)*T))	TRS	00
4003	X14I=X14I+SZ	TRS	00
	X14(K)=SQRTF(X14R**2+X14I**2)	TRS	00
501	RI=RN*(X13(K)/XMAX)	TRS	00
	SI=RI+1.	TRS	00
	XI=RN*(X14(K)/XMAX)	TRS	00
	ZI=XI+1.	TRS	00
	IF(SI-79.) 25,25,26	TRS	00
25	I=SI	TRS	00
	XINE(I)=X	TRS	00
	GO TO 29	TRS	00
26	I=N+1	TRS	00
	XINE(I)=BLANK	TRS	00
29	IF(ZI-79.) 27,27,28	TRS	00
27	II=ZI	TRS	00
	XINE(II)=STAR	TRS	00
	GO TO 75	TRS	00
28	I=N+1	TRS	00
	XINE(I)=BLANK	TRS	00
75	PUNCH 4,(XINE(J),J=1,JMAX)	TRS	00
4	FORMAT (1H*,79A1)	TRS	00
	DO 179 J=1,JMAX	TRS	00
179	XINE(J)=BLANK	TRS	00
110	CONTINUE	TRS	00
	PUNCH 21	TRS	00
21	FORMAT (79(1H@))	TRS	00
	IF(L) 20,23,24	TRS	00
20	PUNCH 22	TRS	00
22	FORMAT (5X,10HMAGNITUDES,1X,9HX.PLOTTED,1X,5HCURVE,1X,3HTHE,1X, 15HFIRST,1X,6HOUTPUT)	TRS	00
	PUNCH 221,(X13(I),I=1,M)	TRS	00
221	FORMAT (8E10.3)	TRS	00
	PUNCH 222	TRS	00
222	FORMAT (5X,10HMAGNITUDES,1X,2HOF,1X,9H*.PLOTTED,1X,5HCURVE,1X,3HFOR 1,1X,3HTHE,1X,5HFIRST,1X,6HOUTPUT)	TRS	00
	PUNCH 221,(X14(I),I=1,M)	TRS	00
	GO TO 120	TRS	00
23	PUNCH 224	TRS	00
224	FORMAT (5X,10HMAGNITUDES,1X,2HOF,1X,9HX.PLOTTED,1X,5HCURVE,1X,3 1HFOR,1X,3HTHE,1X,6HSECOND,1X,6HOUTPUT)	TRS	01
	PUNCH 221,(X13(I),I=1,M)	TRS	01
	PUNCH 226	TRS	01
226	FORMAT (5X,9HMAGNITUDE, 1X,2HOF,1X,9H*.PLOTTED,1X,5HCURVE,1X,3HFOR 11X,3HTHE,1X,6HSECOND,1X,6HOUTPUT)	TRS	01
	PUNCH 221,(X14(I),I=1,M)	TRS	01
	GO TO 120	TRS	01
24	PUNCH 227	TRS	01
227	FORMAT (5X,9HMAGNITUDE, 1X,2HOF,1X,9HX.PLOTTED,1X,5HCURVE, 1 1X,3HTHE,1X,5HTHIRD,1X,6HOUTPUT)	TRS	01


```
PUNCH 221,(?13(I),I=1,M) TRS 011
PUNCH 228 TRS 011
228 FORMAT (5X,9HMAGNITUDE, 1X,2HOF,1X,9H*.PLOTTED,1X,5HCURVE, TRS 011
1 1X,3HTHE,1X,5HTHIRD,1X,6HOUTPUT) TRS 011
120 STOP TRS 011
END TRS 011
*****
```

RESULTS OF SRD ALGORITHM *****

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

1.000 .738 6.174 2.861 2.935 1.000 .738 1.011
.111 0.000

SYSTEM IS UNSTABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000 0.000 0.000 0.000 0.000 0.000 2.426 14.5
55.380 149.041 315.186 541.397 774.815 935.668 958.443 834.5
613.869 378.414 191.788 77.886 24.299 5.421 .820 .0
.005 0.000 0.000 0.000 0.000 0.000 5.000 33.5
143.999 440.806 1073.495 2147.144 3635.502 5281.963 6657.291 7319.0
7035.034 5913.235 4328.874 2746.428 1492.637 686.434 260.606 79.5
18.575 3.149 .373 .030 .001 0.000 0.000 0.0

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000 1.000 70.738 39.674 41.201 0.000 1.000 .7
1.174 .111

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000 0.000 0.000 0.000 0.000 0.000 0.000 33.9
204.310 775.326 2086.578 4412.603 7579.563 10847.420 13099.368 13418.2
11683.396 8594.166 5297.809 2685.035 1090.411 340.188 75.907 11.4
1.138 .070 .002 0.000 0.000 0.000 0.000 70.0
470.120 2015.987 6171.288 15028.935 30060.023 50897.038 73947.491 93202.0
ER F8Z .10246657E+06Z Z
98490.483 82785.303 60604.247 38449.997 20896.926 9610.077 3648.4
1114.127 260.050 44.093 5.233 .420 .021 0.000 0.0

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000 1.000 20.738 12.174 11.851 0.000 1.000 .7
1.174 .111

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27

27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPECTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	0.000	9.7
58.374	221.521	596.166	1260.744	2165.590	099.262	3742.677	3833.7
3338.113	2455.476	1513.659	767.153	311.546	97.196	21.687	3.2
.325	.020	0.000	0.000	0.000	0.000	0.000	20.0
134.320	575.996	1763.225	4293.981	8588.578	14542.011	21127.854	26629.1
29276.163	28140.138	23652.943	17315.499	10985.713	5970.550	2745.736	1042.4
318.322	74.300	12.598	1.495	.120	.006	0.000	0.0

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DENOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DENOMINATOR RESPECTIVELY

0.000	1.000	10.738	6.674	5.981	0.000	1.000	.7
1.174	.111						

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPECTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	0.000	4.8
29.187	110.760	298.082	630.371	1082.794	1549.632	1871.338	1916.8
1669.056	1227.738	756.829	383.576	155.773	48.598	10.843	1.6
.162	.010	0.000	0.000	0.000	0.000	0.000	10.0
67.160	287.998	881.612	2146.990	4294.289	7271.005	10563.926	13314.5
14638.081	14070.069	11826.471	8657.749	5492.856	2985.275	1372.868	521.2
159.161	37.150	6.299	.747	.060	.003	0.000	0.0

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DENOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DENOMINATOR RESPECTIVELY

0.000	1.000	5.738	3.924	3.046	0.000	1.000	.7
1.174	.111						

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPECTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	0.000	2.0
14.593	55.380	149.041	315.186	541.397	774.815	935.668	958.0
834.528	613.869	378.414	191.788	77.886	24.299	5.421	.0
.081	.005	0.000	0.000	0.000	0.000	0.000	5.0
33.580	143.999	440.806	1073.495	2147.144	3635.502	5281.963	6657.0
7319.040	7035.034	5913.235	4328.874	2746.428	1492.637	686.434	260.0
79.580	18.575	3.149	.373	.030	.001	0.000	0.0

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DENOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

1.000 .738 6.174 3.801 2.935 1.000 .738 1.17
.111 0.000

SYSTEM IS UNSTABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000 0.000 0.000 0.000 0.000 2.934 .17.022 65.09
176.198 379.795 667.653 987.386 1241.626 1339.110 1243.341 991.13
677.050 390.885 189.016 74.079 22.979 5.222 .811 .08
.005 0.000 0.000 0.000 0.000 0.000 5.000 34.52
149.795 464.138 1140.132 2298.615 3916.741 5721.843 7243.280 7990.10
7697.951 6477.301 4741.194 3002.374 1626.163 743.370 279.961 84.50
19.464 3.259 .382 .030 .001 0.000 0.000 0.00

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000 1.000 70.738 52.834 41.201 0.000 1.000 .73
1.174 .111

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000 0.000 0.000 0.000 0.000 0.000 41.090 238.32
911.370 2466.819 5317.194 9347.215 13823.475 17382.831 18747.581 17406.80
13875.589 9478.706 5472.400 2646.234 1037.114 321.710 73.110 11.36
1.154 .072 .002 0.000 0.000 0.000 0.000 70.00
ER F8Z .10140592E+06Z Z
483.280 2097.132 6497.939 15961.847 32180.616 54834.377 80105.808
ER F8Z .11186151E+06Z Z
ER F8Z .10777132E+06Z Z
90682.216 66376.729 42033.243 22766.294 10407.181 3919.46
1183.111 272.506 45.637 5.361 .427 .021 0.000 0.00

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000 1.000 20.738 15.934 11.851 0.000 1.000 .73
1.174 .111

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000 0.000 0.000 0.000 0.000 -.005 11.721 63.04

260.291	704.640	1518.969	2670.366	3949.300	4966.301	5356.294	4973.00
3964.411	2708.185	1563.538	756.065	296.318	91.917	20.888	3.00
.329	.020	0.000	0.000	0.000	0.000	0.000	20.00
138.080	599.180	1856.554	4560.528	9194.462	15666.964	22887.373	28973.00
31960.434	30791.805	25909.204	18964.779	12009.497	6504.655	2973.480	1119.00
338.031	77.859	13.039	1.531	.122	.006	0.000	0.00

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000	1.000	10.738	8.554	5.981	0.000	1.000
1.174	.111					

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	5.867	34.00
130.181	352.379	759.566	1335.278	1974.745	2483.230	2678.204	2486.00
1982.220	1354.098	781.770	378.033	148.159	45.958	10.444	1.00
.164	.010	0.000	0.000	0.000	0.000	0.000	10.00
69.040	299.590	928.277	2280.263	4597.231	7833.482	11443.686	14486.00
15980.217	15395.903	12954.601	9482.390	6004.749	3252.327	1486.740	559.00
169.015	38.929	6.519	.765	.061	.003	0.000	0.00

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000	1.000	5.738	4.864	3.046	0.000	1.000	.7
1.174	.111						

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	2.934	17.00
65.094	176.195	379.791	667.649	987.382	1241.623	1339.107	1243.00
991.112	677.049	390.885	189.016	74.079	22.979	5.222	.00
.082	.005	0.000	0.000	0.000	0.000	0.000	5.00
34.520	149.795	464.138	1140.132	2298.615	3916.741	5721.843	7243.00
7990.108	7697.951	6477.301	4741.194	3002.374	1626.163	743.370	279.00
84.507	19.464	3.259	.382	.030	.001	0.000	0.00

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

1.000	.738	6.174	1.051	0.000	1.000	.738	1.00
.111	0.000						

SYSTEM IS UNSTABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPE CTIVELY

0.000	0.000	0.000	0.000	0.000	5.869	34.044	127.00
336.962	703.739	1193.383	1688.589	2016.516	2043.749	1761.999	1283.00

785.457	395.605	160.860	50.663	11.814	1.975	0.231	0.000	31.000
0.000	0.000	0.000	0.000	0.000	0.000	0.000	5.000	31.000
129.903	377.782	872.333	1647.420	2621.025	3556.838	4155.490	4197.000	4197.000
3663.263	2757.257	1773.735	966.192	436.377	159.887	45.584	9.000	9.000
1.476	.158	.011	0.000	0.000	0.000	0.000	0.000	0.000

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR
 4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000	1.000	70.738	14.334	.111	0.000	1.000	0.000	0.000
1.174	.111							

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPECTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	82.180	476.000	0.000
1782.927	4717.555	9852.464	16707.500	23640.389	28231.341	28612.566	24658.000	0.000
17971.965	10996.407	5538.479	2252.042	709.283	165.396	27.661	3.000	0.000
.258	.013	0.000	0.000	0.000	0.000	0.000	70.000	0.000
444.780	1818.651	5288.949	12212.667	23063.882	36694.351	49795.740	58176.000	0.000
58761.126	51285.694	38601.601	24832.302	13526.690	6109.286	2238.425	637.000	0.000
135.241	20.673	2.225	.163	.007	0.000	0.000	0.000	0.000

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000	1.000	20.738	4.934	.111	0.000	1.000	0.000	0.000
1.174	.111							

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPECTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	23.480	136.000	0.000
509.407	1347.872	2814.990	4773.571	6754.397	8066.097	8175.019	7048.000	0.000
5134.847	3141.830	1582.422	643.440	202.652	47.256	7.903	0.000	0.000
.073	.003	0.000	0.000	0.000	0.000	0.000	20.000	0.000
127.080	519.614	1511.128	3489.333	6589.680	10484.100	14227.354	16621.000	0.000
16788.893	14653.055	11029.029	7094.943	3864.768	1745.510	639.550	182.000	0.000
38.640	5.906	.635	.046	.002	0.000	0.000	0.000	0.000

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DINOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DINOMINATOR RESPECTIVELY

0.000	1.000	10.738	3.054	.111	0.000	1.000	0.000	0.000
1.174	.111							

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPECTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	11.740	68.0
254.703	673.936	1407.495	2386.785	3377.198	4033.048	4087.509	3524.0
2567.423	1570.915	791.211	321.720	101.026	23.628	3.951	.4
.036	.001	0.000	0.000	0.000	0.000	0.000	10.0
63.540	259.807	755.564	1744.666	3294.840	5242.050	7113.677	8310.9
8394.446	7326.527	5514.514	3547.471	1932.384	872.755	319.775	91.1
19.320	2.953	.317	.023	.001	0.000	0.000	0.0

ORDER OF SCALAR RETURN DIFFERENCE OF NUMERATOR AND DENOMINATOR

4 4

SCALAR RETURN DIFFERENCE NUMERATOR AND DENOMINATOR RESPECTIVELY

0.000	1.000	5.738	2.114	.111	0.000	1.000	.7
1.174	.111						

SYSTEM IS STABLE

ORDERS OF THE NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT

27 27

NUM. AND DINO. OF THE MODIFIED DIAGONAL ELEMENT RESPECTIVELY

0.000	0.000	0.000	0.000	0.000	0.000	5.870	34.0
127.351	336.968	703.747	1193.392	1688.599	2016.524	2043.754	1762.0
1283.711	785.457	395.605	160.860	50.663	11.814	1.975	.2
.018	0.000	0.000	0.000	0.000	0.000	0.000	5.0
31.770	129.903	377.782	872.333	1647.420	2621.025	3556.838	4155.4
4197.223	3663.263	2757.257	1773.735	966.192	436.377	159.887	45.5
9.660	1.476	.158	.011	0.000	0.000	0.000	0.0

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